

4-BIT CMOS MICROPROCESSOR SLICE

IDT39C01C IDT39C01D IDT39C01E

FEATURES:

- Low-power CEMOS™
 - lcc (max.)Military: 35mACommercial: 30mA

Fast

- IDT39C01C meets 2901C speeds
- IDT39C01D 20% speed upgrade
- IDT39C01E 40% speed upgrade
- · Eight-function ALU
 - Performs addition, two subtraction operations and five logic functions on two source operands
- Expandable
 - Longer word lengths achieved through cascading any number of IDT39C01s
- · Four status flags
 - Carry, overflow, negative and zero
- Pin-compatible and functionally equivalent to all versions of the 2901
- · Available in 40-pin DIP and 44-pin LCC
- Military product available compliant to MIL-STD-883 and DESC Standard Military Drawing (SMD) 5962-88535

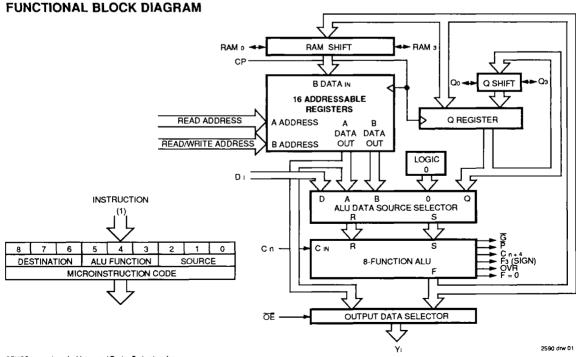
DESCRIPTION:

The IDT39C01s are high-speed, cascadable ALUs which can be used to implement CPUs, peripheral controllers and programmable microprocessors. The IDT39C01's microinstruction flexibility allows for easy emulation of most digital computers.

This extremely low-power yet high-speed ALU consists of a 16-word-by-4-bit dual-port RAM, a high-speed ALU and the required shifting, decoding and multiplexing logic. It is expandable in 4-bit increments, contains a flag output along with three-state data outputs, and can easily use either a ripple carry or full lookahead carry. The nine-bit microinstruction word is organized into three groups of three bits each and selects the ALU destination register, ALU source operands and the ALU function.

The IDT39C01 is fabricated using CEMOS™, a CMOS technology designed for high-performance and high-reliability. It is a pin-compatible, performance-enhanced, functional replacement for all versions of the 2901.

Military grade product is manufactured in compliance with the latest version of MIL-STD-883, Class B.

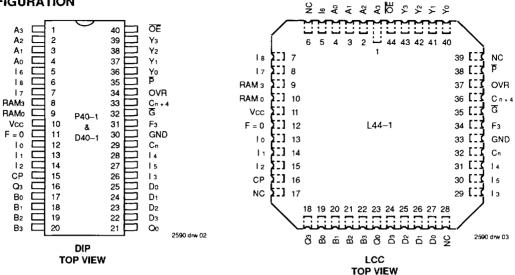


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MILITARY AND COMMERCIAL TEMPERATURE RANGES

APRIL 1990

PIN CONFIGURATION



PIN DESCRIPTIONS

Ao - A3 Bo - B3 Io - I8 Do - D3 Yo - Y3		the ALU will perform 1(3, 4, 5) and what data is to be deposited in the Q Register or the register file 1(6, 7, 8). Four-bit direct data inputs which are the data source for entering external data into the device. Do is the LSB.
lo - la Do - D3	1	displayed through the B port. They also select the location into which new data can be written when the clock goes LOW. Nine instruction control lines which determine what data source will be applied to the ALU I(0, 1, 2), what function the ALU will perform I(3, 4, 5) and what data is to be deposited in the Q Register or the register file I(6, 7, 8).
Do - D3	1	the ALU will perform 1(3, 4, 5) and what data is to be deposited in the Q Register or the register file 1(6, 7, 8). Four-bit direct data inputs which are the data source for entering external data into the device. Do is the LSB.
Yo - Y3	0	Four three-state output lines which when enabled display either the four outputs of the ALU or the data on the A
		port of the register stack. This is determined by the destination code (6, 7, 8).
F3	0	Most significant ALU output bit (sign-bit).
F = 0	0	Open drain output which goes HIGH if the Fo - F3 ALU outputs are all LOW. This indicates that the result of an ALU operation is zero (positive logic).
Cn	Ī	Carry-in to the internal ALU.
Cn+4	0	Carry-out of the internal ALU.
Q3 RAM3	I/O	Bidirectional lines controlled by I(6, 7, 8). Both are three-state output drivers connected to the TTL-compatible CMOS inputs. When the destination code on I(6, 7, 8) indicates an up shift, the three-state outputs are enabled, the MSB of the Q Register is available on the Q3 pin and the MSB of the ALU output is available on the RAM3 pin. When the destination code indicates a down shift, the pins are the data inputs to the MSB of the Q Register and the MSB of the RAM.
Qo RAMo	I/O	Both bidirectional lines function identically to Q3 and RAM3 lines except they are the LSB of the Q Register and RAM.
ŌĒ	I	Output enable on which, when pulled HIGH, the Y outputs are OFF (high impedance). When pulled LOW, the Y outputs are enabled.
G, P	0	Carry generate and carry propagate output of the ALU. These are used to perform a carry lookahead operation.
OVR	0	Overflow. This pin is logically the Exclusive-OR of the carry-in and carry-out of the MSB of the ALU. At the most significant end of the word, this pin indicates that the result of an arithmetic two's complement operation has overflowed into the sign-bit.
СР	I	Clock input. LOW-to-HIGH clock transitions will change the Q Register and the register file outputs. Clock LOW time is internally the write enable time for the 16 x 4 RAM which comprises the master latches of the register file. While the clock is LOW, the slave latches on the RAM outputs are closed, storing the data previously on the RAM outputs. Synchronous MASTER-SLAVE operation of the register file is achieved by this.

ALU SOURCE OPERAND CONTROL

	Microcode				ource ands	
Mnemonic	l2	l1	lo	Octal Code	R	s
AQ	L	L	L	0	Α	Q
AB	L	L	н	1	Α	8
ZQ	L	н	L	2	0	Q
ZB	L.	н	Н	3	0	В
ZA	Н	L	L	4	0	Α
DA	Н	L	н	5	D	Α
DQ	Н	Н	L	6	D	Q
DZ	Н	н	Н	7	D	0

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ALU FUNCTION CONTROL

	Microcode					
Mnemonic	ls	14	l3	Octal Code	ALU Function	Symbol
ADD	L	L	L	0	R Plus S	R+S
SUBR	L	L	н	1	S Minus R	S-R
SUBS	L	Н	L	2	R Minus S	R-S
OR	L	Н	н	3	RORS	RVS
AND	н	L	L	4	RANDS	R۸S
NOTRS	Н	L	н	5	R AND S	R٨S
EXOR	Н	н	L	6	R EX-OR S	R⊽S
EXNOR	Н	н	Н	7	R EX-NOR S	RVS

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DEVICE ARCHITECTURE:

The IDT39C01 CMOS bit-slice microprocessor is configured four bits wide and is cascadable to any number of bits (4, 8, 12, 16, etc.). Key elements which make up this four-bit microprocessor slice are: 1) the register file (16 x 4 dual-port RAM) with shifter; 2) ALU and 3) Q Register and shifter.

REGISTER FILE - RAM data is read from the A port as controlled by the 4-bit A address field input. Data, as defined by the B address field input, can be simultaneously read from the B port of the RAM. This same code can be applied to the A select and B select field with the identical data appearing at both the RAM A port and B port outputs, simultaneously. New data is written into the file (word) defined by the B address field of the RAM when activated by the RAM write enable. The RAM data input field is driven by a 3-input multiplexer that is used to shift the ALU output data (F). It is capable of shifting the data up one position, down one position or not shifting at all. The other inputs to the multiplexer are from the RAM3 and RAMo I/O pins. For a shift up operation, the RAM3 output buffer is enabled and the RAMo multiplexer input is enabled. During a shift down operation, the RAMo output buffer is enabled and the RAM3 multiplexer input is enabled. Four-bit latches hold the RAM data while the clock is LOW, with the A port output and B port output each driving separate latches. The data to be written into the RAM is applied from the ALU F output.

ALU — The ALU can perform three binary arithmetic and five logic operations on the two 4-bit input words S and R. The S input field is driven from a 3-input multiplexer and the R input field is driven from a 2-input multiplexer, with both having an inhibit capability. Both multiplexers are controlled by the lo, l1, l2 inputs. This multiplexer configuration enables the user to select various pairs of the A, B, D, Q and "0" inputs as source operands to the ALU. Microinstruction inputs (I3, I4, I5) are

used to select the ALU function. This high-speed ALU also incorporates a carry-in (Cn) input, carry propagate (\overline{P}) output, carry generate (\overline{G}) output and carry-out (Cn+4) all aimed at accelerating arithmetic operations by the use of carry look ahead logic. The overflow output pin (OVR) will be HIGH when arithmetic operations exceed the two's complement number range. The ALU data outputs (F0, F1, F2, F3) are routed to the RAM, Q Register inputs and the Y outputs under control of the I6, I7, Is control signal inputs. The MSB of the ALU is output as F3 so the user can examine the sign-bit without enabling the three-state outputs. An open drain output, F = 0, is HIGH when F0 = F1 = F2 = F3 = 0 so the user can determine when the ALU output is zero by wire-ORing these outputs together.

Q REGISTER — The Q Register is a separate 4-bit file intended for multiplication and division routines and can also be used as an accumulator or holding register for other types of applications. It is driven from a 3-input multiplexer. In the no-shift mode, the multiplexer enters the ALU data into the Q Register. In either the shift-up or shift-down mode, the multiplexer selects the Q Register data appropriately shifted up or down. The Q shifter has two ports, Qo and Q3, which operate comparably to the RAM shifter. They are controlled by the I6, I7, I8 inputs.

The clock input of the IDT39C01 controls the RAM, Q Register and A and B data latches. When enabled, the data is clocked into the Q Register on the LOW-to-HIGH transition. When the clock is HIGH, the A and B latches are open and pass data that is present at the RAM outputs. When the clock is LOW, the latches are closed and retain the last data entered. When the clock is LOW and RAM EN is enabled, new data will be written into the RAM file defined by the B address field.

ALU DESTINATION CONTROL(1)

	Microcode			RAM Function		Q Register Function		RAM Shifter		Q Shifter			
Mnemonic	ls	17	16	Octal Code	Shift	Load	Shift	Load	Y Output	RAM ₀	RAM3	Qo	Q3
QREG	L	L	L	0	Х	NONE	NONE	F→Q	F	×	×	Х	Х
NOP	L	L	Н	1	Х	NONE	Х	NONE	F	х	х	×	Х
RAMA	L	Н	L	2	NONE	F→B	X	NONE	Α	Х	Х	Х	Х
RAMF	L	Н	Н	3	NONE	F→B	Х	NONE	F	×	Х	X	Х
RAMQD	Н	L	L	4	DOWN	F/2 → B	DOWN	Q/2 → Q	F	Fo	lNз	Qo	INз
RAMD	Н	L	Н	5	DOWN	F/2 → B	х	NONE	F	Fo	lNз	Qo	Х
RAMQU	Н	Н	L	6	UP	2F → B	UP	2Q → Q	F	IN ₀	Fз	INo	Q3
RAMU	Н	Н	Н	7	ÜP	2F → B	х	NONE	F	INo	Fз	Х	Qз

NOTE:

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SOURCE OPERAND AND ALU FUNCTION MATRIX⁽¹⁾

			l2, 1, 0 Octal						
		0	1	2	3	4	5	6	7
Octal	ALU				ALU	Source			
l5, 4, 3	Function	A, Q	A, B	0, Q	0, B	0, A	D, A	D, Q	D, 0
0	Cn = L R Plus S	A + Q	A + B	a	В	A	D + A	D + Q	D
	Cn ≃ H	A+Q+1	A + B + 1	Q+1	B + 1	A + 1	D + A + 1	D+Q+1	D+1
1	Cn = L S Minus R	Q – A – 1	B – A – 1	Q – 1	B – 1	A 1	A - D - 1	Q – D – 1	-D − 1
	Cn = H	Q – A	B – A	Q	В	A	A – D	Q – D	_D
2	Cn = L R Minus S	A – Q – 1	A – B – 1	Q - 1	-B - 1	-A − 1	D - A - 1	D – Q – 1	D – 1
	Cn = H	A – Q	A – B	-Q	-B	-A	D – A	D – Q	D
3	RORS	AVQ	A V B	Q	В	Α	DVA	DVQ	D
4	RANDS	AΛQ	AΛB	0	0	0	DΛA	DΛQ	0
5	R AND S	ĀΛQ	ĀΛB	a.	В	Α	DΛA	DΛQ	0
6	R EX-OR S	AVQ	ΑVΒ	Q	В	Α	D 7 A	DVQ	D
7	R EX-NOR S	Ā∇Q	ΑVΒ	ä	B	Ã	DΫĀ	DVQ	D

5.1

NOTE:

1. + = Plus; - = Minus; Λ = AND; ∇ = EX-OR; ∀ = OR.

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¹ X = Don't care. Electrically, the shift pin is a TTL input internally connected to a three-state output which is in the high-impedance state. B = Register Addressed by B inputs.
UP is toward MSB; DOWN is toward LSB.

E

ALU LOGIC MODE FUNCTIONS

Oc	tal		
l5, 4, 3	l2, 1, 0	Group	Function
4 4 4 4	0 1 5 6	AND	A
3 3 3 3	0 1 5 6	OR	A V Q A V B D V A D V Q
6 6 6	0 1 5 6	EX-OR	A 7 Q A 7 B D 7 A D 7 Q
7 7 7 7	0 1 5 6	EX-NOR	A V Q A V B D V A D V Q
7 7 7 7	2 3 4 7	INVERT	<u>Q</u> <u>В</u> A <u>D</u>
6666	2 3 4 7	PASS	Q B A D
3 3 3	2 3 4 7	PASS	Q B A D
4 4 4 4	2 3 4 7	"ZERO"	0 0 0
5 5 5 5	0 1 5 6	MASK	Ā Λ Q Ā Λ B D Λ A D Λ Q
			2590 tbl 06

ALU ARITHMETIC MODE FUNCTIONS

Oc	tal	Cn	= L	Cn	= H
l5, 4, 3	12, 1, 0	Group	Function	Group	Function
0 0 0	0 1 5 6	ADD	A + Q A + B D + A D + Q	ADD plus one	A+Q+1 A+B+1 D+A+1 D+Q+1
0 0 0	2 3 4 7	PASS	Q B A D	Increment	Q+1 B+1 A+1 D+1
1 1 1 2	2 3 4 7	Decrement	Q - 1 B - 1 A - 1 D - 1	PASS	Ø ∰ ≼ D
2 2 2	2 3 4 7	1's Comp.	-Q - 1 -B - 1 -A - 1 -D - 1	2's Comp. (Negate)	Q # 4 p
1 1 1 1 2 2 2 2	0 1 5 6 0 1 5 6	Subtract (1's Comp)	Q-A-1 B-A-1 A-D-1 Q-D-1 A-Q-1 A-B-1 D-A-1 D-Q-1	Subtract (2's Comp)	Q - A B - D Q - D A - D A - D D - Q

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DEFINITIONS(1)

Po = Ro + So
P1 = R1 + S1
$P_2 = R_2 + S_2$
P3 = R3 + S3
Go = RoSo
G1 = R1S1
G2 = R2S2
G3 = R3S3
C4 = G3 + P3G2 + P3P2G1 + P3P2P1G0 + P3P2P1P0Cn
C3 = G2 + P2G1 + P2P1G0 + P2P1P0Cn

NOTE:

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LOGIC FUNCTIONS FOR \overline{G} , \overline{P} , Cn + 4 AND OVR (1)

l5, 4, 3	Function	P	G	C n + 4	OVR
0	R+S	P3P2P1P0	G3 + P3G2 + P3P2G1 + P3P2P1G0	C4	C3 ∇ C4
1	S-R		Same as R + S equations, but sub	stitute Ri for Ri in definition	s
2	R-S	-	Same as R + S equations, but sub	stitute Si for Si in definition	s
3	RVS	LOW	P3P2P1P0	P3P2P1P0 + Cn	P3P2P1P0 + Cn
4	R۸S	LOW	G3 + G2 + G1 + G0	G3 + G2 + G1 + G0 + Cn	G3 + G2 + G1 + G0 + Cr
5	R۸S	LOW	Same as R V S equation	ons, but substitute Rifor Ri	in definitions
6	R⊽S		Same as RVS equations, but sub	stitute Ri for Ri in definition	s
7	RÝS	G3 + G2 + G1 + G0	G3 + P3G2 + P3P2G1 + P3P2P1P0	G3 + P3G2 + P3P2G1 + P3P2P1P0 (G0 + Cn)	(2)

NOTES:

 $\begin{array}{lll} 1 & += \mathsf{OR}. \\ 2. & & [\bar{\mathsf{P}}_2 + \bar{\mathsf{G}}_2\bar{\mathsf{P}}_1 + \bar{\mathsf{G}}_2\bar{\mathsf{G}}_1\bar{\mathsf{P}}_0 + \bar{\mathsf{G}}_2\bar{\mathsf{G}}_1\bar{\mathsf{G}}_0\mathsf{C}_n] \ \nabla \ [\bar{\mathsf{P}}_3 + \bar{\mathsf{G}}_3\bar{\mathsf{P}}_2 + \bar{\mathsf{G}}_3\bar{\mathsf{G}}_2\bar{\mathsf{P}}_1 + \bar{\mathsf{G}}_3\bar{\mathsf{G}}_2\bar{\mathsf{G}}_1\bar{\mathsf{P}}_0 + \bar{\mathsf{G}}_3\bar{\mathsf{G}}_2\bar{\mathsf{G}}_1\bar{\mathsf{C}}_n] \end{array}$

ABSOLUTE MAXIMUM RATINGS(1)

Symbol	Rating	Com'l.	Mil.	Unit
Vcc	Power Supply Voltage	-0.5 to +7.0	-0.5 to +7.0	٧
VTERM	Terminal Voltage with Respect to Ground	-0.5 to Vcc +0.5	-0.5 to Vcc +0.5	٧
Та	Operating Temperature	0 to +70	-55 to +125	°C
TBIAS	Temperature Under Bias	-55 to +125	-65 to +135	ç
Тѕтс	Storage Temperature	-55 to +125	-65 to +150	ů
Рт	Power Dissipation	1.0	1.0	W
lout	DC Output Current	30	30	mA

CAPACITANCE (TA = +25°C, f = 1.0MHz)

Symbol	Parameter ⁽¹⁾	Conditions	Тур.	Unit
Cin	Input Capacitance	VIN = OV	5	pF
Cour	Output Capacitance	Vout = 0V	7	pF

NOTE:

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NOTE:

1 Stresses greater than those listed under ABSOLUTE MAXIMUM RATINGS may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating

conditions for extended periods may affect reliability. DC ELECTRICAL CHARACTERISTICS

Commercial: TA = 0° C to $+70^{\circ}$ C, Vcc = $5.0V \pm 5\%$; Military: TA = -55° C to $+125^{\circ}$ C, Vcc = $5.0V \pm 10\%$

Symbol	Parameter	Test Co	nditions	Min.	Typ. ⁽³⁾	Max.	Unit
Lін	Input HIGH Current (All Inputs)	Vcc = Max. Vin = Vcc		-	0.1	5	μА
l II.	Input LOW Current (All Inputs)	Vcc = Max. Vin = GND			-0.1	-5	μА
Vон	Output High Voltage	Vcc = Min.	IOH = -1.0mA (MIL.)	2.4	4.3	_	V
		Vin = VIH or VIL	IOH = -1.6mA (COM'L.)	2.4	4.3	-	
Vol	Output Low Voltage	Vcc = Min.	IOL = 16mA (MIL.)	_	0.3	0.5	٧
		VIN = VIH or VIL	IOL = 20mA (COM'L.)		0.3	0.5	<u></u> .
Vін	Input HIGH Level	Guaranteed Logic HIGH	Levei ⁽¹⁾	2.0	_	_	٧
VIL	Input LOW Level	Guaranteed Logic LOW I	Level ⁽¹⁾	_	_	0.8	٧
loz	Output Leakage Current	Vcc = Max.	Vout = 0V	_	-0.1	-10	μА
			Vout = Vcc (Max.)	_	0.1	10	
los	Output Short Circuit Current	VCC = Max. VOUT = 0V ⁽²⁾		-30	-	-	mA

NOTES:

1. These input levels provide zero noise immunity and should only be static tested in a noise-free environment.

2. Not more than one output should be shorted at a time. Duration of the short circuit test should not exceed one second.

3. Vcc = 5.0V at TA +25°C.

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^{1.} This parameter is sampled and not 100% tested.

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DC ELECTRICAL CHARACTERISTICS (Cont'd.)

Commercial: TA = 0°C to +70°C, Vcc = $5.0V \pm 5\%$; Military: TA = -55°C to +125°C, Vcc = $5.0V \pm 10\%$

VLC = 0.2V; VHC = VCC -0.2V

Symbol	Parameter	Test Condi	tions ⁽¹⁾		Min.	Typ. (3)	Max.	Unit
Іссан	Quiescent Power Supply Current	Vcc = Max.			ļ	0.5	5.0	mA
	CP = H (CMOS Inputs)	VHC ≤ VIH, VIL ≤ VLC						
		fcP = 0, CP = H						
ICCOL	Quiescent Power Supply Current	Vcc = Max.			1	0.5	5.0	mA
	CP = L (CMOS Inputs)	VHC ≤ VIH, VIL ≤ VLC						
		fcP = 0, CP = L						
Ісст	Quiescent Input Power Supply ⁽⁴⁾	Vcc = Max., VIH = 3.4V			1	0.3	0.5	mA/
	Current (per Input @ TTL High)	fCP = 0						Input
ICCD	Dynamic Power Supply Current	Vcc = Max.		MIL.	1	1.5	2.5	mA/
		VHC ≤ VIH, VIL ≤ VLC		COM'L.	-	1.0	2.0	MHz
		Outputs Open, OE = L						
Icc	Total Power Supply Current ⁽⁵⁾	Vcc = Max	1DT39C01C	MIL.	1	_	30	mA.
		Outputs Open, OE = L	fcP = 10MHz	COM'L.	1		25]
		CP = 50 % Duty cycle	IDT39C01D	MIL.	-	-	35	ļ
		VHC ≤ VIH, VIL ≤ VLC	fcP = 15MHz	COM'L.	1	_	30	
		50% Data Duty Cycle	IDT39C01E	MIL.	1	-	40	
			fcP = 17.5MHz	COM'L.	-	_	35	
		Vcc = Max	IDT39C01C	MIL.	-	-	35	
		Outputs Open, \overline{OE} = L	fcP = 10MHz	COM'L.	-	-	30	
		CP = 50 % Duty Cycle	IDT39C01D	MIL.	1	_	40	
		VIH = 3.4V, VIL = 0.4V	fcp = 15MHz	COM'L.		_	35	
		50% Data Duty Cycle	IDT39C01E	MIL.	-	_	45	
			fcP = 17.5MHz	COM'L.	-	_	40	

NOTES:

These input levels should only be static tested in a noise-free environment.

- 2. Not more than one output should be shorted at a time. Duration of the short circuit test should not exceed one second.
- 3. Vcc = 5 0V at TA +25°C
- 4. IccT is derived by measuring the total current with all the inputs tied together at 3.4V, subtracting out IccOH, then dividing by the total number of inputs.
- 5. Total Supply Current is the sum of the Quiescent current and the Dynamic current (at either CMOS or TTL input levels). For all conditions, the Total Supply Current can be calculated by using the following equation:

ICC = ICCOH (CDH) +ICCOL (1 - CDH) + ICCT (NT x DH) + ICCD (fCP)

CDH = Clock duty cycle high period

DH = Data duty cycle TTL high period (VIN = 3.4V)

NT = Number of dynamic inputs driven at TTL levels

fcp = Clock input frequency

CMOS TESTING CONSIDERATIONS

There are certain testing considerations which must be taken into account when testing high-speed CMOS devices in an automatic environment. These are:

- Proper decoupling at the test head is necessary. Placement of the capacitor set and the value of capacitors used is critical in reducing the potential erroneous failures resulting from large Vcc current changes. Capacitor lead length must be short and as close to the DUT power pins as possible.
- All input pins should be connected to a voltage potential during testing. If left floating, the device may begin to oscillate causing improper device operation and possible latchup.
- 3) Definition of input levels is very important. Since many inputs may change coincidentally, significant noise at the device pins may cause the VIL and VIH levels not to be met until the noise has settled. To allow for this testing/board induced noise, IDT recommends using VIL ≤ 0V and VIH ≥ 3V for AC tests.
- 4) Device grounding is extremely important for proper device testing. The use of multi-layer performance boards with radial decoupling between power and ground planes is required. The ground plane must be sustained from the performance board to the DUT interface board. All unused interconnect pins must be properly connected to the ground pin. Heavy gauge stranded wire should be used for power wiring and twisted pairs are recommended to minimize inductance.

AC ELECTRICAL CHARACTERISTICS IDT39C01C

(Military and Commercial Temperature Ranges)

The tables below specify the guaranteed performance of the IDT39C01C over the -55° C to $+125^{\circ}$ C and 0° C to $+70^{\circ}$ C temperature ranges. VCC is specified at $5V \pm 10\%$ for military temperature range and $5V \pm 5\%$ for commercial temperature range. All times are in nanoseconds and are measured at the 1.5V signal level. The inputs switch between 0V and 3V with signal transition rates of 1V per nanosecond. All outputs have maximum DC current loads.

CYCLE TIME AND CLOCK CHARACTERISTICS

	Mil.	Com'l.	Unit
Read-Modity-Write Cycle (from selection of A, B registers to end of cycle)	32	31	ns
Maximum Clock Frequency to shift Q (50% duty cycle, I = 432 or 632)	31	32	MHz
Minimum Clock LOW Time	15	15	ns
Minimum Clock HIGH Time	15	15	ns
Minimum Clock Period	32	31	ns

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COMBINATIONAL PROPAGATION DELAYS(1) CL = 50pF

	1							To	Outp	out	_		_				
	1	Y		F3	С	n + 4	7	5, P	F	= 0	o	VR		AMo AM3	ĺ	Q0 Q3	
From Input	Mil.	Com'l.	Mil.	Com'l.	Unit												
A, B Address	48	40	48	40	48	40	44	37	48	40	48	40	48	40	-	-	ns
D	37	30	37	30	37	30	34	30	40	38	37	30	37	30	-	1	ns
Cn	25	22	25	22	21	20	-	-	28	25	25	22	28	25	-	+	ns
lo, 1, 2	40	35	40	35	40	35	44	37	44	37	40	35	40	35	ı	_	ns
13, 4, 5	40	35	40	35	40	35	40	35	40	38	40	35	40	35	-	_	ns
l6, 7, 8	29	25	_	_	_		1	~	ı	_	-	-	29	26	29	26	ns
A Bypass ALU (I = 2XX)	40	35	-	-	-	_	-	_	1	-	1	1	1	-	1	-	ns
Clock ✓	40	35	40	35	40	35	40	35	40	35	40	35	40	35	33	28	ns

SET-UP AND HOLD TIMES RELATIVE TO CLOCK (CP INPUT)

	CP:	CP: tpwL												
		p Time e H → L		l Time H → L		p Time I L → H	Hold After							
Input	Mil.	Com'l.	Mil.	Com'i.	MII.	Com'l.	Mil.	Com'i.	Unit					
A, B Source Address	15	15	2	1 (3)	30,15 +	· tPWL ⁽⁴⁾	2	1	ns					
B Destination Address	15	15	_	Do not c	2	1	ns							
D	_ (1)	-	_	_	25	25	0	0	ns					
Cn		_	_	_	20	20	0	0	ns					
10, 1, 2	_	_	-	-	30	30	0	0	ns					
13, 4, 5	-	-	_	_	30	30	0	0	ns					
l 6, 7, 8	10	10	Do not change (2)				0	0	ns					
RAM0,3, Q0,3		<u> </u>	_	-	12	12	0	0	ns					

NOTES:

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- 1. A dash indicates a propagation delay or set-up time constraint does not exist.
- 2. Certain signals must be stable during the entire clock LOW time to avoid erroneous operation.
- Source addresses must be stable prior to the H → L transition to allow time to access the source data before the latches close. The A address may then
 be changed. The B address could be changed if it is not a destination; i.e., if data is not being written back into the RAM. Normally A and B are not changed
 during the clock LOW time.
- 4. The set-up time prior to the clock L → H transition is to allow time for data to be accessed, passed through the ALU and returned to the RAM. It includes all the time from stable A and B addresses to the clock L → H transition, regardless of when the H → L transition occurs.

OUTPUT ENABLE/DISABLE TIMES

(CL = 5pF, measured to 0.5V

change of Vour in nanoseconds)

		Ena	ble	Dis	able
Input	Output	Mil.	Com'l.	Mil.	Com'l.
ŌĒ	Y	25	23	25	23

AC ELECTRICAL CHARACTERISTICS IDT39C01D

(Military and Commercial Temperature Ranges)

The tables below specify the guaranteed performance of the IDT39C01D over the -55° C to $+125^{\circ}$ C and 0° C to $+70^{\circ}$ C temperature ranges. VCC is specified at $5V \pm 10\%$ for military temperature range and $5V \pm 5\%$ for commercial temperature range. All times are in nanoseconds and are measured at the 1.5V signal level. The inputs switch between 0V and 3V with signal transition rates of 1V per nanosecond. All outputs have maximum DC current loads.

CYCLE TIME AND CLOCK CHARACTERISTICS

	Mil.	Com'l.	Unit
Read-Modify-Write Cycle (from selection of A, B registers to end of cycle)	27	23	ns
Maximum Clock Frequency to shift Q (50% duty cycle, I = 432 or 632)	37	43	MHz
Minimum Clock LOW Time	13	11	ns
Minimum Clock HIGH Time	13	11	ns
Minimum Clock Period	27	23	ns

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COMBINATIONAL PROPAGATION DELAYS(1) CL = 50pF

								To	Outp	out							
	Y		F3		Cn +4		G, P		F=0		OVR		RAMo RAM3		Qo Q3		
From Input	Mil.	Com'l.	Mil.	Com'i.	Mil.	Com'l.	Mił.	Com'l.	Mil.	Com'l.	Mil.	Com'l.	Mil.	Com'l.	Mil.	Com'l.	Unit
A, B Address	33	30	33	30	33	30	33	30	33	30	33	30	33	30			ns
D	24	21	23	20	23	20	21	20	25	24	24	21	25	22	-	-	ns
Cn	18	17	17	16	14	14	-	-	19	18	17	16	19	18	-	-	ns
lo, 1, 2	28	26	27	25	26	24	28	24	29	25	27	24	27	25	-	-	ns
13, 4, 5	27	26	27	24	26	24	26	24	27	26	26	24	27	26			ns
l 6, 7, 8	18	16	-	-	-		-	_	-	_	_	-	21	21	21	21	ns
A Bypass ALU (I = 2XX)	26	24	-	_	-	-		_	-	-	-	_	ı	1	-	-	ns
Clock /	27	24	26	23	26	23	25	23	27	24	26	24	27	24	20	19	ns

5.1

SET-UP AND HOLD TIMES RELATIVE TO CLOCK (CP INPUT)

	CP:	CP: 1PWL											
		p Time e H → L	Hold Time After H → L			Time L → H	Hold After						
input	Mil.	Com'l.	Mit.	Com'l.	Mil.	Com'i.	Mil.	Com'l.	Unit				
A, B Source Address	11	10	0	0 (3)	24, 11 + tPWL ⁽⁴⁾	21, 10 + tPWL ⁽⁴⁾	2	1	ns				
B Destination Address	11	10		Do not c	2	1	ns						
D	_ (1)		_	_	16	16	0	0	ns				
Cn	_	_		-	13	13	0	0	ns				
10, 1, 2	_	—	_	T -	19	19	0	0	ns				
13, 4, 5	_	-	-	_	19	19	0	0	ns				
l6, 7, 8	7	7		Do not c	hange ⁽²⁾		0	0	ns				
RAM0,3, Q0,3	_		_		9	9	0	0	ns				

NOTES:

- 1. A dash indicates a propagation delay or set-up time constraint does not exist.
- 2. Certain signals must be stable during the entire clock LOW time to avoid erroneous operation.
- 3. Source addresses must be stable prior to the H -> L transition to allow time to access the source data before the latches close. The A address may then be changed. The Bladdress could be changed if it is not a destination; i.e., if data is not being written back into the RAM. Normally A and B are not changed during the clock LOW time.
- 4. The set-up time prior to the clock L → H transition is to allow time for data to be accessed, passed through the ALU and returned to the RAM. It includes all the time from stable A and B addresses to the clock $L \to H$ transition, regardless of when the $H \to L$ transition occurs.

OUTPUT ENABLE/DISABLE TIMES

(CL = 5pF, measured to 0.5V

change of Vout in nanoseconds)

		Ena	ble	Dis	able
Input	Output	Mil.	Com'l.	Mil.	Com'l.
ŌĒ	Υ	16	14	18	16

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AC ELECTRICAL CHARACTERISTICS IDT39C01E

(Military and Commercial Temperature Ranges)

The tables below specify the guaranteed performance of the IDT39C01E over the -55° C to $+125^{\circ}$ C and 0° C to $+70^{\circ}$ C temperature ranges. VCC is specified at $5V \pm 10\%$ for military temperature range and $5V \pm 5\%$ for commercial temperature range. All times are in nanoseconds and are measured at the 1.5V signal level. The inputs switch between 0V and 3V with signal transition rates of 1V per nanosecond. All outputs have maximum DC current loads.

CYCLE TIME AND CLOCK CHARACTERISTICS

	Mil.	Com'l.	Unit
Read-Modify-Write Cycle (from selection of A, B registers to end of cycle)	21	20	ns
Maximum Clock Frequency to shift Q (50% duty cycle, I = 432 or 632)	46	. 50	MHz
Minimum Clock LOW Time	10	8	ns
Minimum Clock HIGH Time	10	8	ns
Minimum Clock Period	21	20	ns

2500 512

COMBINATIONAL PROPAGATION DELAYS (1) CL = 50pF

		To Output															
		Y		F3	С	n + 4	7	5, P	F	= 0	c	VR		AMo AM3		Q ₀ Q ₃	
From Input	Mil.	Com'l.	Mil.	Com'l.	Mil.	Com'l.	Mil.	Com'l.	Mil.	Com'l.	Mil.	Com'l.	Mil.	Com'l.	Mil.	Com'l.	Unit
A, B Address	26	22	26	22	26	22	26	21	29	25	26	22	26	22	_	-	ns
D	18	16	17	15	17	15	16	15	22	20	18	16	19	16	-	_	ns
Cn	13	13	13	12	10	10	_	_	16	15	13	12	14	13	-	_	ns
10, 1, 2	21	20	20	19	19	18	21	18	25	21	20	18	20	19	_	_	ns
13, 4, 5	20	20	20	18	19	18	19	18	23	23	19	18	20	20	_	-	ns
l 6, 7, 8	13	12	-	-	_	-	-	-	-	-	_	-	16	16	16	16	ns
A Bypass ALU (I = 2XX)	26	24	_	_	-	-	_	-	_	_	_	_	-	-	-	_	ns
Clock /	20	18	19	17	19	17	19	17	25	22	19	18	20	18	15	15	ns

SET-UP AND HOLD TIMES RELATIVE TO CLOCK (CP INPUT)

CP:	7	<u> </u>	—— tpv	VL ———			-	
		Hold Time After H → L						
Mil.	Com'l.	Mil.	Com'l.	Mil.	Com'l.	Mil.	Com'i.	Unit
8	7	0	0 (3)	18, 8 + tPWL ⁽⁴⁾	15, 7 + tPWL ⁽⁴⁾	2	1	ns
8	7	Do not change (2)				2	1	ns
_ (1)		_		12	12	0	0	ns
_	-	-	_	10	10	0	0	ns
_	-	_	_	14	14	0	0	ns
_	-	_	-	14	14	0	0	ns
5	5		Do not c	hange ⁽²⁾		0	0	ns
_	_	_	_	9	9	0	0	ns
	Set-u Before Mil. 8 8 (1) 	Set-up Time Before H → L Mil. Com'l. 8 7 8 7 - (1) 5 5	Set-up Time Hold After Before H → L After Mil. Com'l. Mil. 8 7 0 8 7 - - (1)	CP: Set-up Time Before H → L Hold Time After H → L Mil. Com'l. Mil. Com'l. 8 7 0 0 (3) 8 7 Do not c — (1) — — — — — — — — — — — — — — — — — — —	Set-up Time Before H → L Hold Time After H → L Set-up Before Mil. Com'l. Mil. Com'l. Mil. 8 7 0 0 (3) 18, 8 + tPWL (4) 8 7 Do not change (2) — (1) — — 12 — — — 10 — — — 14 — — — 14 5 5 Do not change (2)	CP: Set-up Time Before H → L After H → L Set-up Time Mil. Com'l. Mil. Com'l. 8 7 0 0 (3) 18, 8 + 15, 7 + 19WL (4) 15, 7 + 19WL (4) 8 7 Do not change (2) 12 12 — (1) — — — — — — — — — — — — 10 10 10 — — — — — — — — — — — — — 14 14 14 — — — — — — — — — — — — — — — — — — —	$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$

NOTES:

- 1. A dash indicates a propagation delay or set-up time constraint does not exist.
- 2. Certain signals must be stable during the entire clock LOW time to avoid erroneous operation.
- 3. Source addresses must be stable prior to the H \rightarrow L transition to allow time to access the source data before the latches close. The A address may then be changed. The B address could be changed if it is not a destination; i.e., if data is not being written back into the RAM. Normally A and B are not changed during the clock LOW time.
- 4. The set-up time prior to the clock L → H transition is to allow time for data to be accessed, passed through the ALU and returned to the RAM. It includes all the time from stable A and B addresses to the clock $L \to H$ transition, regardless of when the $H \to L$ transition occurs.

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OUTPUT ENABLE/DISABLE TIMES

(CL = 5pF, measured to 0.5V

change of Vout in nanoseconds)

		Enable		Dis	able
Input	Output	Mil.	Com'l.	Mil.	Com'l.
ŌĒ	Y	14	10	12	12
					0500 #10

AC TEST CONDITIONS

Input Pulse Levels Input Rise/Fall Times Input Timing Reference Levels Output Reference Levels	GND to 3.0V 1V/ns 1.5V 1.5V
Output Load	See Figure 4

Test	Switch
Open Drain	Closed
Disable Low	
Enable Low	
All other Tests	Open

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INPUT/OUTPUT INTERFACE CIRCUIT

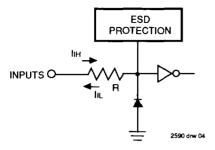


Figure 1. Input Structure (All Inputs)

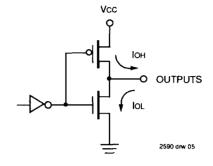


Figure 2. Outputs Structure (All Outputs Except F=0)

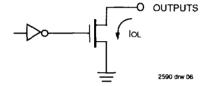


Figure 3. Output Structure (F=0 Only)

TEST CIRCUIT LOAD

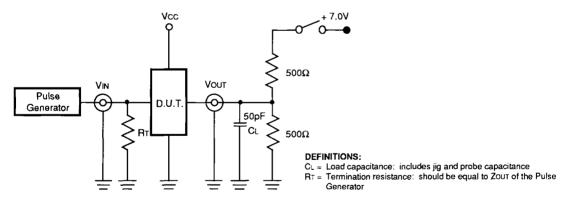


Figure 4. Switching Test Circuits

5.1

ORDERING INFORMATION

