

Delphi NC30 Series Non-Isolated Point of Load DC/DC Power Modules: 12Vin, 0.9V-5Vout, 30A

The Delphi NC30 Series, 12V input, single output, non-isolated point of load DC/DC converters are the latest offering from a world leader in power systems technology and manufacturing — Delta Electronics, Inc. The NC30 series operates from a 12V nominal input, provides up to 30A of power in a vertical or horizontal mounted through-hole package and the output can be resistor- or voltage-trimmed from 0.9Vdc to 5.0Vdc. NC30 series has built-in current sharing control and multiple NC30/NC40 series modules could be paralleled together to provide even higher output currents. NC30 series provides a very cost effective point of load solution. With creative design technology and optimization of component placement, these converters possess outstanding electrical and thermal performance, as well as extremely high reliability under highly stressful operating conditions.

FEATURES

- High Efficiency:94% @ 12Vin, 5V/30A out
- Voltage and resistor-based trim
- No minimum load required
- Output voltage programmable from
 0.9Vdc to 5.0Vdc via external resistors
- Fixed frequency operation
 - Input UVLO, output OVP, OTP, OCP, SCP
- Remote ON/OFF (default: positive)
- Power good output signal
- Output voltage sense
- ISO 9001, TL 9000, ISO 14001, QS 9000, OHSAS 18001 certified manufacturing facility
- UL/cUL 60950 (US & Canada) Recognized, and TUV (EN60950) Certified
- CE mark meets 73/23/EEC and 93/68/EEC directives

OPTIONS

- Vertical or horizontal versions
- Negative On/Off logic

APPLICATIONS

- DataCom
- Distributed power architectures
- Servers and workstations
- LAN/WAN applications
- Data processing applications



TECHNICAL SPECIFICATIONS

 $(T_A=25^{\circ}C, airflow rate=400LFM, V_{in}=12Vdc, nominal Vout unless otherwise noted.)$

PARAMETER	NOTES and CONDITIONS	NC12S0A0V30			
		Min.	Тур.	Max.	Units
ABSOLUTE MAXIMUM RATINGS					
Input Voltage				14	Vdc
Operating Temperature	D (0		50	°C
Storage Temperature	Refer to Figure 36 for the measuring point	-40	NIA	125	°C
Input/Output Isolation Voltage	Non-isolated		NA		V
INPUT CHARACTERISTICS		10.2	12	12.0	V
Operating Input Voltage Input Under-Voltage Lockout		10.2	12	13.8	V
Turn-On Voltage Threshold			9.0		V
Turn-Off Voltage Threshold			8.3		V
Lockout Hysteresis Voltage			0.7		V
Maximum Input Current	100% Load, 10.2Vin, 5Vout		0.7	15.6	A
No-Load Input Current	100 /0 2000, 10.2 viii, 0 vout		160	10.0	mA
Off Converter Input Current			100		mA
Input Reflected-Ripple Current	Refer to Figure 35		150		mA
Input Voltage Ripple Rejection	120 Hz		55		dB
OUTPUT CHARACTERISTICS			30		30
Output Voltage Adjustment Range		0.9		5.0	V
Output Voltage Set Point	Vin=12V, Io=Io,max, Ta=25°€, 1% trim resistors	-3.0		+3.0	%
Output Voltage Regulation	, , , , , , , , , , , , , , , , , , , ,	0.0		0.0	,,,
Over Load	lo=lo,min to lo,max	-1.0		+1.0	%
Over Line	Vin=Vin,min to Vin,max	-0.2		+0.2	%
Output Voltage Ripple and Noise	5Hz to 20MHz bandwidth				
Peak-to-Peak	Full Load, 1µF ceramic, 10µF tantalum			50	mV
RMS	Full Load, 1µF ceramic, 10µF tantalum			15	mV
Output Current Range	The state of the s	0		30	Α
Output Voltage Over-shoot at Start-up	Vin=12V, Turn ON			1	%
Output Voltage Under-shoot at Power-Off	Vin=12V, Turn OFF			100	mV
Output DC Current-Limit Inception	,	36			Α
Output Short-Circuit Current (Hiccup mode)		36			Α
DYNAMIC CHARACTERISTICS					
Out Dynamic Load Response	12Vin, 10μF Tan & 1μF Ceramic load cap, 10A/μs				
Positive Step Change in Output Current	50% lo,max to 75% lo,max		75		mV
Negative Step Change in Output Current	75% lo,max to 50% lo,max		75		mV
Setting Time	Settling to be within regulation band (+/- 3.0%)			150	μs
Turn-On Transient	lo=lo.max				
Start-Up Time, From On/Off Control	Vin=12V, Vo=10% of Vo,set, Ta=25 $^{\circ}$ C			10	ms
Start-Up Time, From Input	Vo=10% of Vo,set, Ta=25°C			30	ms
Minimum Output Startup Capacitive Load	Ex: Two OSCON $6.3V/680\mu F$ (ESR $13m\Omega$ max each)	1360			
Maximum Output Startup Capacitive Load	Full load; ESR ≥10mΩ			5440	μF
Minimum Input Capacitance	Ex: OSCON 16V/270 μ F (ESR 18m Ω max)	270			μF
EFFICIENCY					
Vo=0.9V	Vin=12V, Io=30A		78		%
Vo=1.2V	Vin=12V, Io=30A		82		%
Vo=1.5V	Vin=12V, Io=30A		85		%
Vo=1.8V	Vin=12V, Io=30A		87		%
Vo=2.5V	Vin=12V, Io=30A		90		%
Vo=3.3V	Vin=12V, Io=30A		92		%
Vo=5.0V	Vin=12V, Io=30A		94		%
FEATURE CHARACTERISTICS			000		171.1
Switching Frequency	Decilia hada fisharadh a dhababa		300		KHz
ON/OFF Control	Positive logic (internally pulled high)	6.1		\	* *
Logic High	Module On (or leave the pin open)	2.4		Vin,max	V
Logic Low	Module Off	-0.2		0.8	V
Remote Sense Range				0.4	V
GENERAL SPECIFICATIONS			1.00		Maria
MTBF			1.69		M hours
Weight Over-Temperature Shutdown	Auto restort refer to Fig. 269.44 for the recognition of the		36		grams
Over-Temperature Shutdown	Auto restart, refer to Fig. 36&41 for the measuring point		130		°C

ELECTRICAL CHARACTERISTICS CURVES

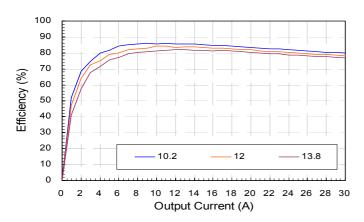


Figure 1: Converter efficiency vs. output current (0.9V output voltage)

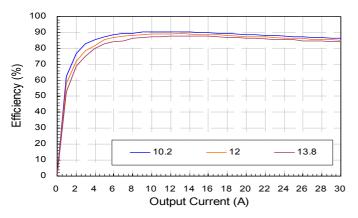


Figure 3: Converter efficiency vs. output current (1.5V output voltage)

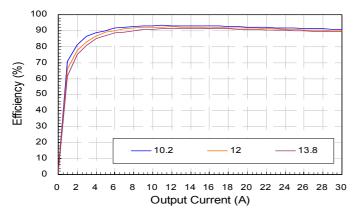


Figure 5: Converter efficiency vs. output current (2.5V output voltage)

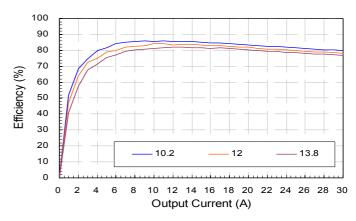


Figure 2: Converter efficiency vs. output current (1.2V output voltage)

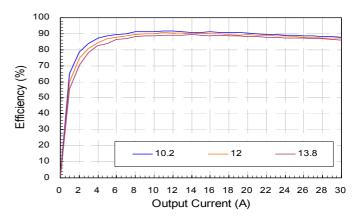


Figure 4: Converter efficiency vs. output current (1.8V output voltage)

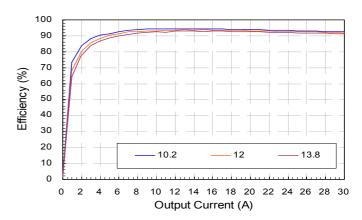


Figure 6: Converter efficiency vs. output current (3.3V output voltage)

ELECTRICAL CHARACTERISTICS CURVES (CON.)

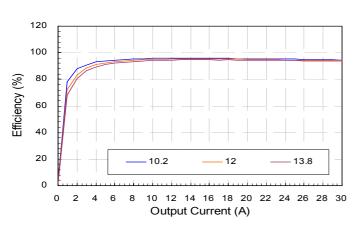


Figure 7: Converter efficiency vs. output current (5.0V output voltage)

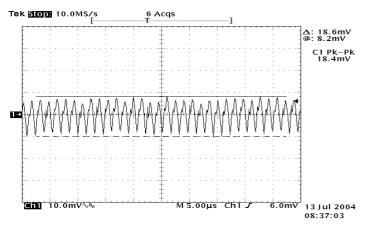


Figure 9: Output ripple & noise at 12Vin, 1.2V/30A out

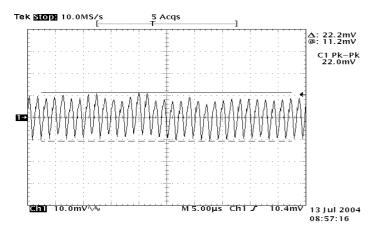


Figure 11: Output ripple & noise at 12Vin, 1.8V/30A out

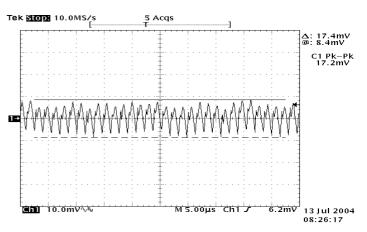


Figure 8: Output ripple & noise at 12Vin, 0.9V/30A out

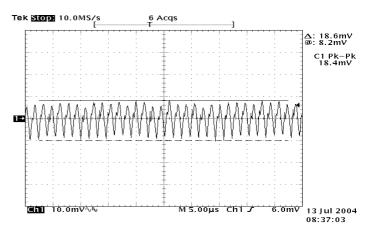


Figure 10: Output ripple & noise at 12Vin, 1.5V/30A out

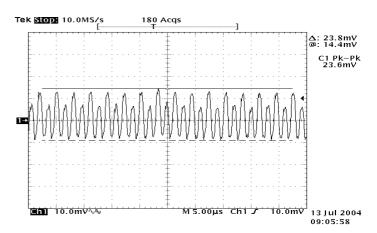


Figure 12: Output ripple & noise at 12Vin, 2.5V/30A out

ELECTRICAL CHARACTERISTICS CURVES (CON.)

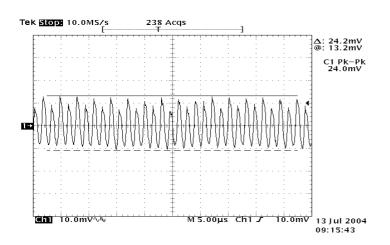


Figure 13: Output ripple & noise at 12Vin, 3.3V/30A out

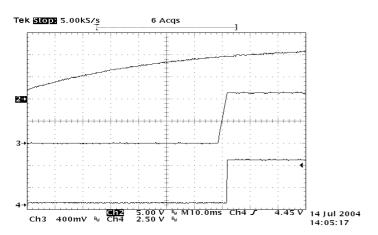


Figure 15: Turn on delay time at Vin On/Off, 0.9V/30A out Ch2:Vin Ch3:Vout Ch4:PWRGD

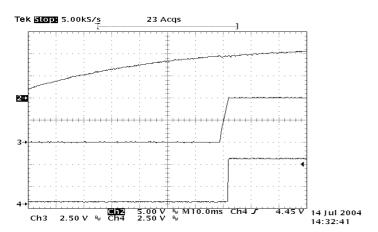


Figure 17: Turn on delay time at 12vin, 5.0V/30A out Ch2:Vin Ch3:Vout Ch4:PWRGD

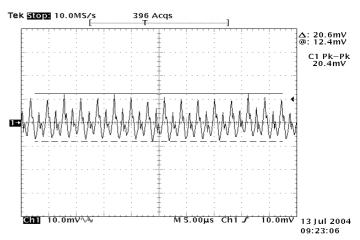


Figure 14: Output ripple & noise at 12Vin, 5.0V/30A out

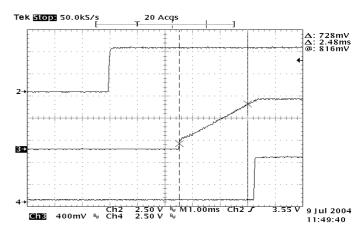


Figure 16:Turn on delay time at Remote On/Off, 0.9V/30A out Ch2:ENABLE Ch3:Vout Ch4:PWRGD

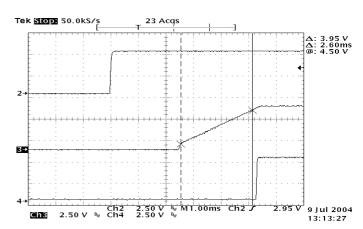


Figure 18: Turn on delay time at Remote On/Off, 5.0V/30A out Ch2: ENABLE Ch3:Vout Ch4:PWRGD

ELECTRICAL CHARACTERISTICS CURVES (CON.)

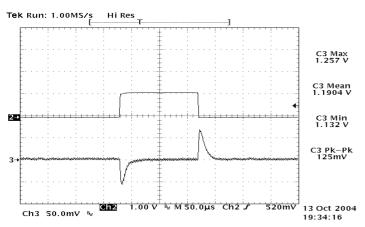


Figure 19: Typical transient response to step load change at $10A/\mu S$ from 75% to 50% of Io, max at 12Vin, 1.2V out (Cout = $1\mu S$ ceramic, $10\mu F$ tantalum)

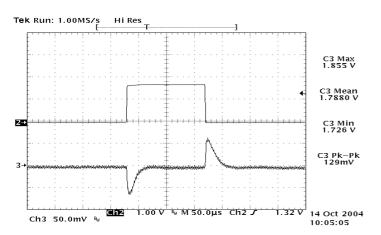


Figure 21: Typical transient response to step load change at $10A/\mu S$ from 75% to 50% of lo, max at 12Vin, 1.8V out (Cout = $1\mu S$ ceramic, $10\mu F$ tantalum)

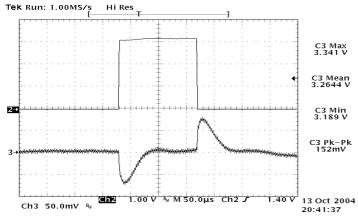


Figure 23: Typical transient response to step load change at $10A/\mu S$ from 75% to 50% of lo, max at 12Vin, 3.3V out (Cout = $1\mu S$ ceramic, $10\mu F$ tantalum)

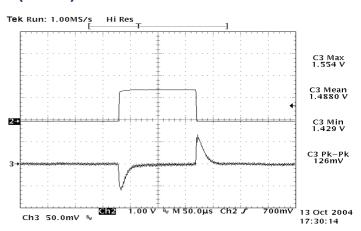


Figure 20: Typical transient response to step load change at $10A/\mu S$ from 75% to 50% of Io, max at 12Vin, 1.5V out (Cout = $1\mu S$) (Cout

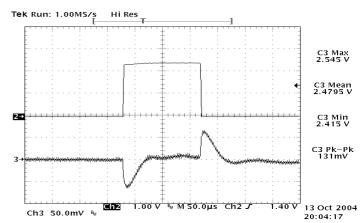


Figure 22: Typical transient response to step load change at $10A/\mu S$ from 75% to 50% of Io, max at 12Vin, 2.5V out (Cout = $1\mu S$) (Cout = $1\mu S$) (Cout = $1\mu S$) (Cout = $1\mu S$)

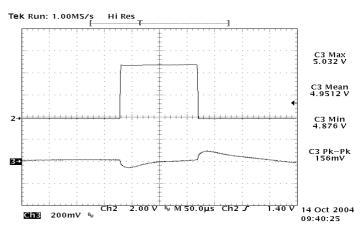


Figure 24: Typical transient response to step load change at $10A/\mu S$ from 75% to 50% of lo, max at 12Vin, 5.0V out (Cout = $1\mu S$) (Cout = $1\mu S$) (Cout = $1\mu S$) (Cout = $1\mu S$)

DESIGN CONSIDERATIONS

The NC30 is designed using two-phase synchronous buck topology. Block diagram of the converter is shown in Figure 25. The output can be trimmed in the range of 0.9Vdc to 5.0Vdc by a resistor from trim pin to ground. A remote sense function is provided and it is able to compensate for a drop from the output of converter to point of load.

The converter can be turned ON/OFF by remote control. Positive on/off (ENABLE pin) logic implies that the converter DC output is enabled when this signal is driven high (greater than 2.4V) or floating and disabled when the signal is driven low (below 0.8V). Negative on/off logic is optional and could also be ordered.

The converter provides an open collector signal called Power Good. The power good signal is pulled low when output is not within ±10% of Vout or Enable is OFF.

The converter can protect itself by entering hiccup mode against over current and short circuit condition. Also, the converter will shut down when an over voltage protection is detected.

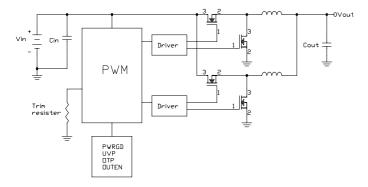


Figure 25: Block Diagram

Safety Considerations

It is recommended that the user to provide two 12A very fast-acting type fuses (Little fuse R451 012) in parallel in the input line for safety.

FEATURES DESCRIPTIONS

ENABLE (On/Off)

The ENABLE (on/off) input allows external circuitry to put the NC converter into a low power dissipation (sleep) mode. Positive (active-high) ENABLE is available as standard.

Positive ENABLE (active-high) units of the NC series are turned on if the ENABLE pin is high or floating. Pulling the pin low will turn off the unit. With the active high function, the output is guaranteed to turn on if the ENABLE pin is driven above 2.4V. The output will turn off if the ENABLE pin voltage is pulled below .8V.

The ENABLE input can be driven in a variety of ways as shown in Figures 26, 27 and 28. If the ENABLE signal comes from the primary side of the circuit, the ENABLE can be driven through either a bipolar signal transistor (Figure 26) or a logic gate (Figure 27). If the enable signal comes from the secondary side, then an opto-coupler or other isolation devices must be used to bring the signal across the voltage isolation (please see Figure 28).

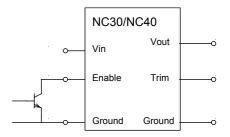


Figure 26: Enable Input drive circuit for NC series

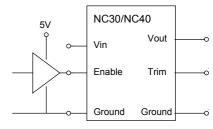


Figure 27: Enable input drive circuit using logic gate.

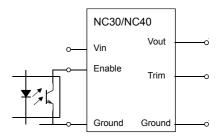


Figure 28: Enable input drive circuit example with isolation.

FEATURES DESCRIPTIONS (CON.)

Input Under-Voltage Lockout

The input under-voltage lockout prevents the converter from being damaged while operating when the input voltage is too low. The lockout occurs between 7.7V to 8.6V.

Over-Current and Short-Circuit Protection

The NC series modules have non-latching over-current and short-circuit protection circuitry. When over current condition occurs, the module goes into the non-latching hiccup mode. When the over-current condition is removed, the module will resume normal operation.

An over current condition is detected by measuring the voltage drop across the high-side MOSFET. The voltage drop across the MOSFET is also a function of the MOSFET's Rds(on). Rds(on) is affected by temperature, therefore ambient temperature will affect the current limit inception point.

The unit will not be damaged in an over current condition because it will be protected by the over temperature protection.

Remote Sense

The NC30/NC40 provide Vo remote sensing to achieve proper regulation at the load points and reduce effects of distribution losses on output line. In the event of an open remote sense line, the module shall maintain local sense regulation through an internal resistor. The module shall correct for a total of 0.4V of loss. The remote sense connects as shown in Figures 29.

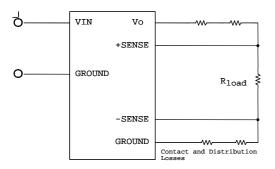


Figure 29: Circuit configuration for remote sense

Over Temperature Protection (OTP)

To provide additional over-temperature protection in a fault condition, the unit is equipped with a non-latching thermal shutdown circuit. The shutdown circuit engages when the temperature of monitored component exceeds approximately 130°C. The unit will cycle on and off while the fault condition exists. The unit will recover from shutdown when the cause of the over temperature condition is removed.

Over Voltage Protection (OVP)

The converter will shut down when an output over voltage is detected. Once the OVP condition is detected, the controller will stop all PWM outputs and will turn on low-side MOSFET driver to prevent any damage to load.

Current Sharing (optional)

The parallel operation of multiple converters is available with the NC30/NC40 (option code B). The converters will current share to be within +/- 10% of each other. In addition to connect the I-Share pin together for the current sharing operation, the remote sense lines of the paralleled units must be connected at the same point for proper operation. Also, units are intended to be turned on/enabled at the same time. Hot plugging is not recommended. The current sharing diagram show in Figure 30.

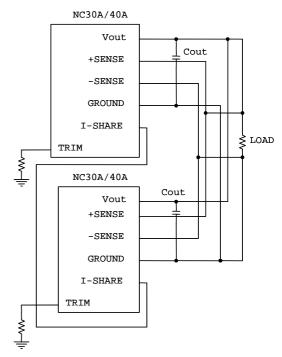


Figure 30: NC30/NC40 Current Sharing Diagram

FEATURES DESCRIPTIONS (CON.)

Output Voltage Programming

The output voltage of the NC series is trimmable by connecting an external resistor between the trim pin and output ground as shown Figure 31 and the typical trim resistor values are shown in Figure 32. The output can also be set by an external voltage connected to trim pin as shown in Figure 32.

The NC30A/40A module has a trim range of 0.9V to 5.0V. A plot of trim behavior is shown in Figure 33

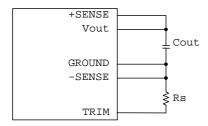


Figure 31: Trimming Output Voltage

The NC30/NC40 modules have a trim range of 0.9V to 5.0V. The trim resistor equation for the them is :

Rs
$$(k\Omega) = \frac{12.69 - Vout}{Vout - 0.9}$$

Vout is the desired voltage setpoint, Rs is the trim resistance between TRIM and Ground, Rs values should not be less than 1.8 k_Ω

0.44.1/-14	D-(O)
Output Voltage	$Rs(\Omega)$
+0.9 V	OPEN
+1.2 V	38.3K
+1.5 V	18.7K
+1.8 V	12.1K
+2.5 V	6.34K
+3.3 V	3.92K
+5.0 V	1.87K

Figure 32: Typical trim resistor values

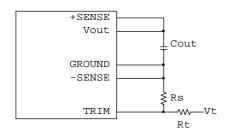


Figure 33: Output voltage trim with voltage source DS_NC12S30A_02072007

To use voltage trim, the trim equation for the NC30 is (please refer to Fig. 33) :

$$Rt(k\Omega) = \frac{Rs(13.1Vt + Vout - 12.69)}{0.9Rs - Vout(Rs + 1) + 12.69}$$

Vout is the desired output voltage Vt is the external trim voltage Rs is the resistance between Trim and Ground (in $K\Omega$) Rt is the resistor to be defined with the trim voltage (in $K\Omega$)

Below is an example about using this voltage trim equation:

Example:

If Vt = 1.25V, desired Vout = 2.5V and Rs = 1 $k\Omega$

$$Rt(k\Omega) = \frac{Rs(13.1Vt + Vout - 12.69)}{0.9Rs - Vout(Rs + 1) + 12.69} = 0.72k\Omega$$

Power Good

The converter provides an open collector signal called Power Good. This output pin uses positive logic and is open collector. This power good output is able to sink 5mA and set high when the output is within $\pm 10\%$ of output set point. The power good signal is pulled low when output is not within $\pm 10\%$ of Vout or Enable is OFF.

Output Capacitance

There is no output capacitor on the NC series modules. Hence, an external output capacitor is required for stable operation. For NC30 modules, two external $6.3V/680\mu F$ output low ESR capacitors in parallel (for example, OSCON) are required for stable operation.

It is important to places these low ESR capacitors as close to the load as possible in order to get improved dynamic response and better voltage regulation, especially when the load current is large. Several of these low ESR capacitors could be used together to further lower the ESR.

Please refer to individual datasheet for the maximum allowed start-up load capacitance for each NC series as it is varied between series.

FEATURES DESCRIPTIONS (CON.)

Voltage Margining

Output voltage margining can be implemented in the NC30/NC40 modules by connecting a resistor, R margin-up, from the Trim pin to the ground pin for margining up the output voltage. Also, the output voltage can be adjusted lower by connecting a resistor, Rmargin-down, from the Trim pin to the output pin. Figure 34 shows the circuit configuration for output voltage margining adjustment.

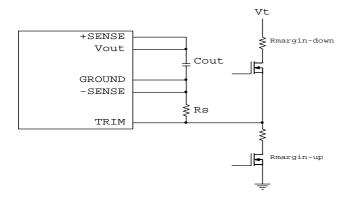
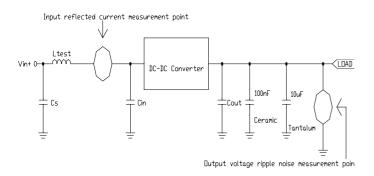


Figure 34: Circuit configuration for output voltage margining

Reflected Ripple Current and Output Ripple and Noise Measurement

The measurement set-up outlined in Figure 35 has been used for both input reflected/ terminal ripple current and output voltage ripple and noise measurements on NC series converters.



Cs=270uF*1 Ltest=1.4uH Cin=270uF*1 Cout=680uF*2

Figure 35: Input reflected ripple/ capacitor ripple current and output voltage ripple and noise measurement setup for NC30

THERMAL CONSIDERATION

Thermal management is an important part of the system design. To ensure proper, reliable operation, sufficient cooling of the power module is needed over the entire temperature range of the module. Convection cooling is usually the dominant mode of heat transfer.

Hence, the choice of equipment to characterize the thermal performance of the power module is a wind tunnel.

Thermal Testing Setup

Delta's DC/DC power modules are characterized in heated vertical wind tunnels that simulate the thermal environments encountered in most electronics equipment. This type of equipment commonly uses vertically mounted circuit cards in cabinet racks in which the power modules are mounted.

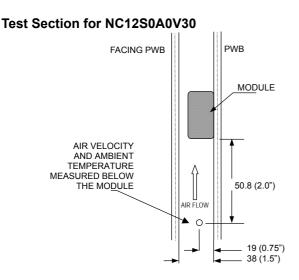
The following figure shows the wind tunnel characterization setup. The power module is mounted on a test PWB and is vertically positioned within the wind tunnel.

Thermal Derating

Heat can be removed by increasing airflow over the module. To enhance system reliability, the power module should always be operated below the maximum operating temperature. If the temperature exceeds the maximum module temperature, reliability of the unit may be affected.

The maximum acceptable temperature measured at the thermal reference point is 125° °C. This is shown in Figure 36 & 41.

THERMAL CURVES (NC12S0A0V30)



Note: Wind Tunnel Test Setup Figure Dimensions are in millimeters and (Inches)

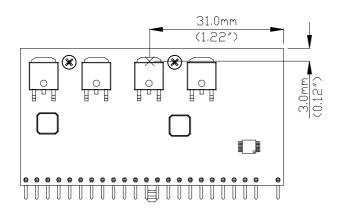


Figure 36: Temperature measurement location * The allowed maximum hot spot temperature is defined at 125 $\mathcal C$

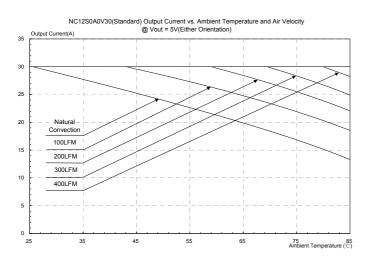
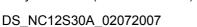


Figure 37: Output current vs. ambient temperature and air velocity @ Vout=5V(Either Orientation)



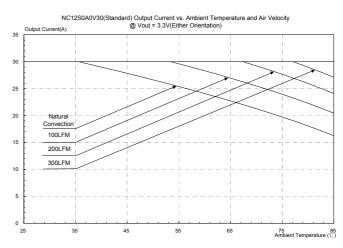


Figure 38: Output current vs. ambient temperature and air velocity @ Vout=3.3V(Either Orientation)

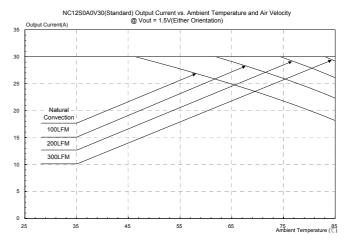


Figure 39: Output current vs. ambient temperature and air velocity @ Vout=1.5V(Either Orientation)

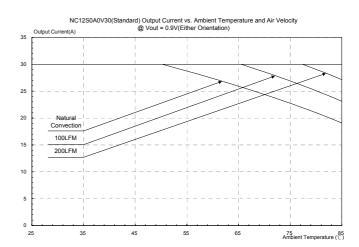
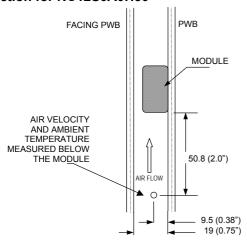


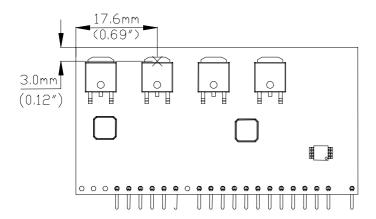
Figure 40: Output current vs. ambient temperature and air velocity @ Vout=0.9V(Either Orientation)

THERMAL CURVES (NC12S0A0H30)

Test Section for NC12S0A0H30



Note: Wind Tunnel Test Setup Figure Dimensions are in millimeters and (Inches)



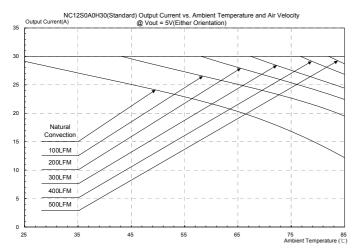


Figure 42: Output current vs. ambient temperature and air velocity @ Vout=5V(Either Orientation)

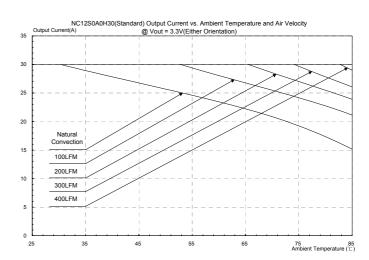


Figure 43: Output current vs. ambient temperature and air velocity @ Vout=3.3V(Either Orientation)

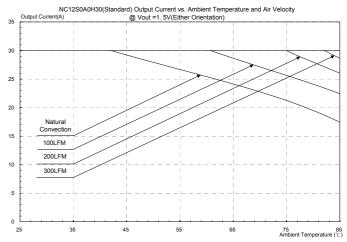


Figure 44: Output current vs. ambient temperature and air velocity @ Vout=1.5V(Either Orientation)

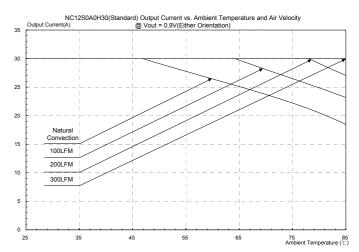
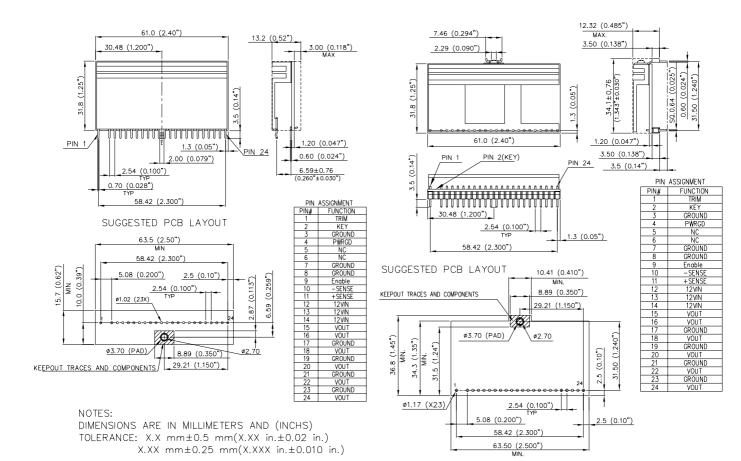


Figure 45: Output current vs. ambient temperature and air velocity @ Vout=0.9V(Either Orientation)

MECHANICAL DRAWING

VERTICAL

HORIZONTAL





NC	12	S	0A0	V	30	Р	N	F	Α	
Product	Input	Number of	Output	Mounting	Output	ON/OFF	Pin Length		Option Code	
Series	Voltage	outputs	Voltage	Woulding	Current	Logic				
NC-	12-	S- Single	0A0-	H- Horizontal	30- 30A	P- Positive	R- 0.118"	F- RoHS 6/6	A- Standard	
Non-isolated	10.2~13.8V	output	programmable	V- Vertical		N- Negative	N- 0.140"	(Lead Free)	Functions	
Converter										

MODEL LIST

Model Name	Packaging	Input Voltage	Oltade Olitolit Voltade Olitolit Clirrent		Efficiency 12Vin @ 100% load
NC12S0A0V30PNFA	Vertical	10.2 ~ 13.8Vdc	0.9 V ~ 5.0Vdc	30A	94% (5.0V)
NC12S0A0H30PNFA	Horizontal	10.2 ~ 13.8Vdc	0.9 V ~ 5.0Vdc	30A	94% (5.0V)

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