

# 4A, 1.5V Fixed Linear Regulator

# Description

The CS-5204-2 linear regulator provides 4A at 1.5V with an accuracy of ±2%.

The fast loop response and low dropout voltage make this regulator ideal for GTL bus termination where low voltage operation and good transient response are important.

The circuit is designed to operate with dropout voltages as low as 1V

depending on the output current level. The maximum quiescent current is only 10mA at full load.

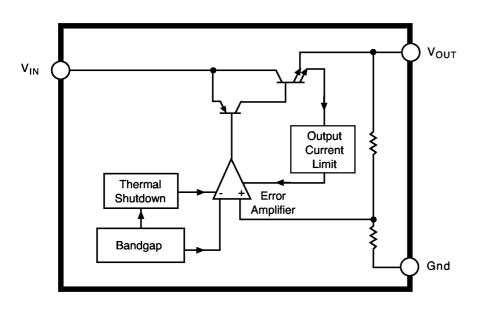
The regulator is fully protected against overload conditions with protection circuitry for Safe Operating Area (SOA), overcurrent and thermal shutdown.

The CS-5204-2 is available in TO-220 and surface mount D<sup>2</sup> packages.

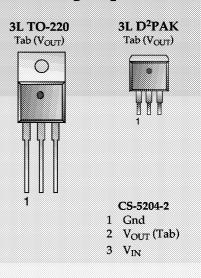
# **Features**

- Output Current to 4A
- Output Voltage Trimmed to ± 2%
- Dropout Voltage 1,10V @ 4A
- Fast Transient Response
- Fault Protection Circuitry
  Thermal Shutdown
  Overcurrent Protection
  Safe Area Protection

### **Block Diagram**



# **Package Options**





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## **Absolute Maximum Ratings**

Supply Voltage, V <sub>CC</sub>	17V
Operating Temperature Range	
Junction Temperature	150°C
Storage Temperature Range	
Lead Temperature Soldering	
Wave Solder (through hole styles only)	10 sec. max, 260°C peak
Reflow (SMD styles only)	

# Electrical Characteristics: $C_{IN} = 10 \mu F$ , $C_{OUT} = 22 \mu F$ Tantalum, $V_{IN} - V_{OUT} = 3V$ , $V_{IN} \le 10V$ , $0^{\circ}C \le T_A \le 70^{\circ}C$ , $T_J \le +150^{\circ}C$ , unless otherwise specified, $I_{full\ load} = 4A$ .

PARAMETER	TEST CONDITIONS		TV F		
■ CS-5204-2					
Output Voltage (Notes 1 and 2)	$V_{IN} - V_{OUT} = 1.5V;$ $0 \le I_{OUT} \le 4A$	1.47 (-2%)	1.50	1.53 (+2%)	V
Line Regulation	$1.5V \le V_{IN} - V_{OUT} \le 6V$ ; $I_{OUT} = 10$ mA		0.04	0.20	%
Load Regulation (Notes 1 and 2)	$V_{IN}$ - $V_{OUT}$ = 1.5V; 10mA $\leq$ $I_{OUT}$ $\leq$ 4A		0.05	0.4	%
Dropout Voltage (Note 3)	I <sub>OUT</sub> =4A		1.1	1.2	V
Current Limit	$V_{IN}$ - $V_{OUT}$ =3V; $T_J \ge 25$ °C $V_{IN}$ - $V_{OUT}$ =15V	4.5	8.5 2.5		A A
Quiescent Current	$V_{IN} \le 9V$ ; $I_{OUT} = 10 \text{mA}$		5.0	10.0	mA
Thermal Regulation	30ms pulse; T <sub>A</sub> =25°C		0.003		%/W
Ripple Rejection	f=120Hz; I <sub>OUT</sub> =4A		75		dB
Temperature Stability			0.5		%
RMS Output Noise (%V <sub>OUT</sub> )	10Hz≤f≤10kHz		0.003		$%V_{OUT}$
Thermal Shutdown		150	180		°C
Thermal Shutdown Hysteresis			25		°C

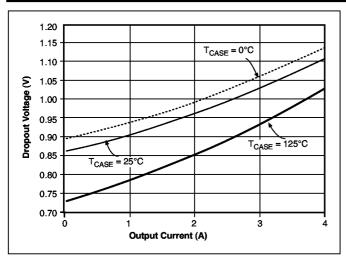
Note 1: Load regulation and output voltage are measured at a constant junction temperature by low duty cycle pulse testing. Changes in output voltage due to thermal gradients or temperature changes must be taken into account separately.

Note 3: Dropout voltage is a measurement of the minimum input/output differential at full load.

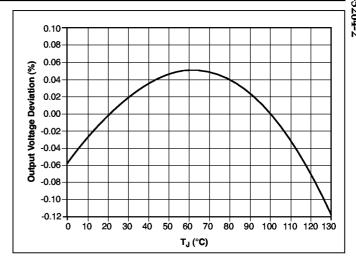
Package Pin Description							
PA		10.5 5 (0.156)	FUNCTION				
D <sup>2</sup> PAK	3L TO-220	-					
1	1	Gnd	Ground connection.				
2	2	V <sub>OUT</sub>	Regulated output voltage (case).				
3	3	$V_{ m IN}$	Input voltage.				

Note 2: Specifications apply for an external Kelvin sense connection at a point on the output pin 1/4" from the bottom of the package.

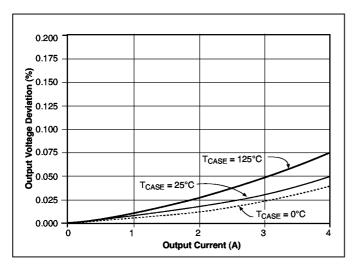
# **Typical Performance Characteristics**



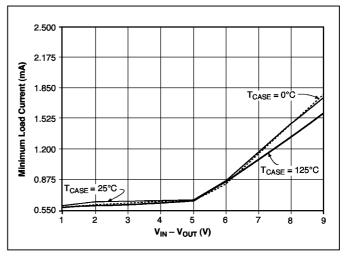
Dropout Voltage vs. Output Current



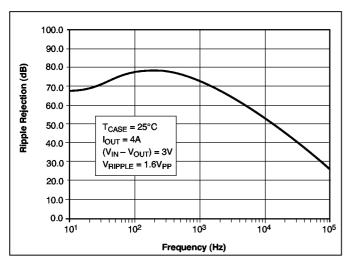
Reference Voltage vs. Temperature



Load Regulation vs. Output Current



**Minimum Load Current** 



Ripple Rejection vs. Frequency

## **Applications Information**

The CS-5204-2 linear regulator provides fixed 1.5V voltage at currents up to 4A. The regulator is protected against short circuit, and includes thermal shutdown and safe area protection (SOA) circuitry. The SOA protection circuitry decreases the maximum available output current as the input-output differential voltage increases.

The CS-5204-2 has a composite PNP-NPN output transistor and requires an output capacitor for stability. A detailed procedure for selecting this capacitor is included in the Stability Considerations section.

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The output or compensation capacitor helps determine three main characteristics of a linear regulator: start-up delay, load transient response and loop stability.

The capacitor value and type is based on cost, availability, size and temperature constraints. A tantalum or aluminum electrolytic capacitor is best, since a film or ceramic capacitor with almost zero ESR can cause instability. The aluminum electrolytic capacitor is the least expensive solution. However, when the circuit operates at low temperatures, both the value and ESR of the capacitor will vary considerably. The capacitor manufacturers' data sheet provides this information.

A  $22\mu F$  tantalum capacitor will work for most applications, but with high current regulators such as the CS-5204-2 the transient response and stability improve with higher values of capacitor. The majority of applications for this regulator involve large changes in load current so the output capacitor must supply the instantaneous load current. The ESR of the output capacitor causes an immediate drop in output voltage given by:

$$\Delta V = \Delta I \times ESR$$

For microprocessor applications it is customary to use an output capacitor network consisting of several tantalum and ceramic capacitors in parallel. This reduces the overall ESR and reduces the instantaneous output voltage drop under load transient conditions. The output capacitor network should be as close as possible to the load for the best results.

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When large external capacitors are used with a linear regulator it is sometimes necessary to add protection diodes. If the input voltage of the regulator gets shorted, the output capacitor will discharge into the output of the regulator. The discharge current depends on the value of the capacitor, the output voltage and the rate at which Vin drops. In the CS-5204-2 linear regulator, the discharge path is through a large junction and protection diodes are not usually needed. If the regulator is used with large values of output capacitance and the input voltage is instantaneously shorted to ground, damage can occur. In this case, a diode connected as shown in Figure 2 is recommended.

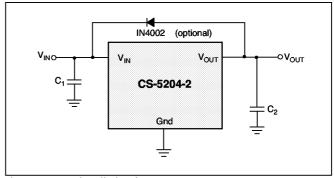


Figure 2. Protection diode scheme.

#### Output Vallage Sensing

Since the CS-5204-2 is a three terminal regulator, it is not possible to provide true remote load sensing. Load regulation is limited by the resistance of the conductors connecting the regulator to the load. For best results the regulator should be connected as shown in Figure 3.

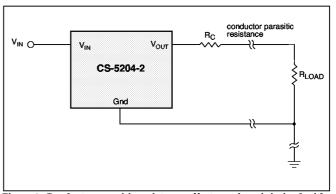


Figure 3. Conductor parasitic resistance effects can be minimized with the above grounding scheme.

#### Calculating Fower Dissipation and Fleat Sink, Requirements

The CS-5204-2 linear regulator includes thermal shutdown and safe operating area circuitry to protect the device. High power regulators such as these usually operate at high junction temperatures so it is important to calculate the power dissipation and junction temperatures accurately to ensure that an adequate heat sink is used.

The case is connected to  $V_{OUT}$  on the CS-5204-2, and electrical isolation may be required for some applications. Thermal compound should always be used with high current regulators such as these.

The thermal characteristics of an IC depend on the following four factors:

- 1. Maximum Ambient Temperature T<sub>A</sub> (°C)
- 2. Power dissipation P<sub>D</sub> (Watts)
- 3. Maximum junction temperature T<sub>I</sub> (°C)
- 4. Thermal resistance junction to ambient  $R_{\Theta IA}$  (C/W)

### **Applications Information: continued**

These four are related by the equation

$$T_{I} = T_{A} + P_{D} \times R_{\Theta I A} \tag{1}$$

The maximum ambient temperature and the power dissipation are determined by the design while the maximum junction temperature and the thermal resistance depend on the manufacturer and the package type.

The maximum power dissipation for a regulator is:

$$P_{D(max)} = \{V_{IN(max)} - V_{OUT(min)}\}I_{OUT(max)} + V_{IN(max)}I_{Q}$$
 (2)

#### where

V<sub>IN(max)</sub> is the maximum input voltage,

V<sub>OUT(min)</sub> is the minimum output voltage,

 $I_{OUT(max)}$  is the maximum output current, for the application

 $I_O$  is the maximum quiescent current at  $I_{OUT}$ (max).

A heat sink effectively increases the surface area of the package to improve the flow of heat away from the IC and into the surrounding air.

Each material in the heat flow path between the IC and the outside environment has a thermal resistance. Like series electrical resistances, these resistances are summed to determine  $R_{\Theta JA}$ , the total thermal resistance between the junction and the surrounding air.

- 1. Thermal Resistance of the junction to case,  $R_{\Theta IC}({}^{\circ}C/W)$
- 2. Thermal Resistance of the case to Heat Sink,  $R_{\Theta CS}$  (°C/W)
- Thermal Resistance of the Heat Sink to the ambient air, R<sub>⊕SA</sub> (°C/W)

These are connected by the equation:

$$R_{\Theta JA} = R_{\Theta JC} + R_{\Theta CS} + R_{\Theta SA}$$
 (3)

The value for  $R_{\Theta JA}$  is calculated using equation (3) and the result can be substituted in equation (1).

 $R_{\Theta JC}$  is 1.6°C/Watt for the CS-5204-2. For a high current regulator such as the CS-5204-2 the majority of the heat is generated in the power transistor section. The value for  $R_{\Theta SA}$  depends on the heat sink type, while  $R_{\Theta CS}$  depends on factors such as package type, heat sink interface (is an insulator and thermal grease used?), and the contact area between the heat sink and the package. Once these calculations are complete, the maximum permissible value of  $R_{\Theta JA}$  can be calculated and the proper heat sink selected. For further discussion on heat sink selection, see application note "Thermal Management for Linear Regulators."

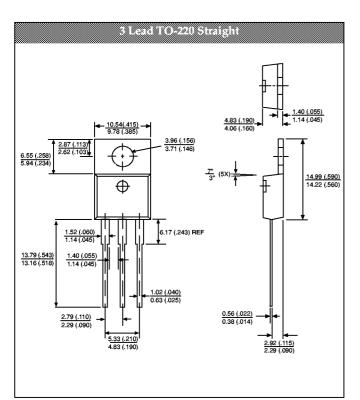
# **Package Specification**

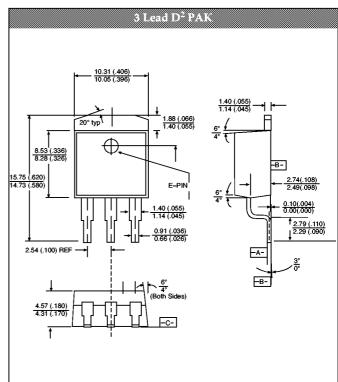
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Thermal	Data	3L TO-220	3L D²PAK				
$R_{\Theta JC}$	typ	1.6	1.6	°C/W			
$R_{\Theta JA}$	typ	50	10 - 50*	°C/W			

\*Depending on thermal properties of substrate.  $R_{\Theta JA} = R_{\Theta JC} + R_{\Theta CA}$ 





Part Number	Туре	Description
CS-5204-2T3	4A, 1.5V output	3 L TO-220 Straight
CS-5204-2DP3	4A, 1.5V output	3 L D <sup>2</sup> PAK

Cherry Semiconductor Corporation reserves the right to make changes to the specifications without notice. Please contact Cherry Semiconductor Corporation for the latest available information.