ST-NXP Wireless

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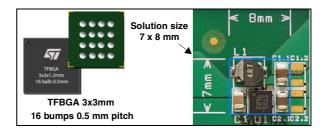


STw4141

Single-coil dual-output step-down DC/DC converter for digital base band and multimedia processor supply

Features

- Single coil dual output switching converter for digital core supply & digital I/Os supply
 - Digital I/O supply:V_{OUT1} @ 200 mA
 - CPU CORE supply:V_{OUT2} @ 400 mA
- Wide range of fixed output voltage configurations available
- High efficiency synchronous step down converter with up to 92 % for the entire device
- Size and cost optimized application board (7x8 mm, height 1.2mm) three capacitors and only one inductor necessary for both outputs
- 2.7 V to 5.5 V battery input range
- ±100mV output voltage accuracy full range in PWM (Including Line and Load Transients)
- 900 kHz fixed frequency PWM operation
- PFM mode operation at light load current
- PWM/PFM switch can be done automatically or forced by setting external pins (AUTO and MODE/SYNC)
- MODE/SYNC input pin for external clock synchronization from 600 kHz to 1.5 MHz
- VSEL input pin for V_{OUT2}/V_{OUT2(red.)} selection
- Ultra low shutdown current (Iq<1 µA)
- Short circuit and thermal shutdown protections



Applications

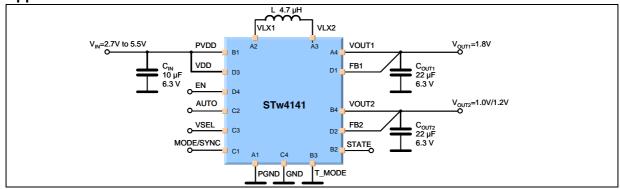
- Mobile phones
- PDAs and hand held terminals
- Portable media players
- Digital still camera
- WLAN and Bluetooth applications

Description

The STw4141 is a single coil dual output synchronous step down DC/DC converter that requires only four standard external components. It operates at a fixed 900 kHz switching frequency in PWM mode. The device can operate in PFM mode to maintain high efficiency over the full range of output currents.

The STw4141 application requires a very small PCB area and offers a very efficient, accurate, space and cost saving solution to fulfill the requirements of digital baseband or multimedia processor supply (CORE & I/O).

Application test circuit



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STw4141 pinout STw4141

1 STw4141 pinout

Figure 1. Pin assignment in TFBGA 3x3 mm - 16 bumps 0.5 mm pitch

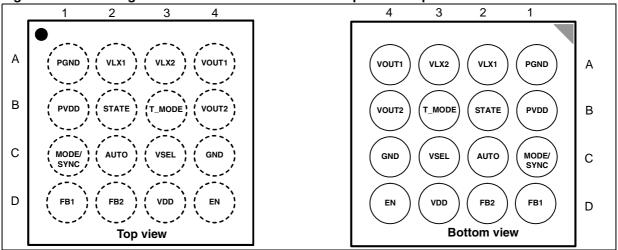


Table 1. STw4141 pin description

Pin	Symbol	Description
A1	PGND	Power ground
B1	PVDD	Power supply voltage
C1	MODE/SYNC	MODE/SYNC = High to forced PWM mode MODE/SYNC = Low to forced PFM mode MODE/SYNC = 600 kHz - 1.5 MHz external clock synchronization in PWM
D1	FB1	Feedback 1
A2	VLX1	External inductor connection pin 1
B2	STATE	Output STATE pin allow the user to monitor operation mode of the product STATE = High - PFM mode STATE = Low - PWM mode If not used must be left unconnected.
C2	AUTO	PWM/PFM automatic switch control pin AUTO = High - PWM/PFM mode automatic switch ENABLED AUTO = Low - PWM/PFM mode automatic switch DISABLED PWM/PFM mode controlled by MODE/SYNC pin)
D2	FB2	Feedback 2
A3	VLX2	External inductor connection pin 2
B3	T_MODE	Input signal for test mode selection. This pin must be connected to GND.
C3	VSEL	Voltage selection input VSEL = High - VOUT1 = 1.8V, VOUT2 = 1.2V (valid for STA1) VSEL = Low - VOUT1 = 1.8V, VOUT2 = 1.0V (valid for STA1) (For other voltage options see <i>Table 1: STw4141 ordering information</i>)
D3	VDD	Signal supply voltage
A4	VOUT1	Output voltage 1

STw4141 STw4141 pinout

Table 1. STw4141 pin description (continued)

Pin	Symbol	Description
B4	VOUT2	Output voltage 2
C4	GND	Signal ground
D4	EN	Enable Input: EN = Low - Device in shutdown mode, EN = High - Enable device This pin must be connected either to VDD or GND.

PGND pinThis is the ground pin related to power signal. This pin should be connected to the board ground

plane by short and wide track or multiply vias to reduce impedance and EMI.

GND pins This is the ground pin related to analog signal.

PVDD pin This pin is designed to provide power to the device. This path leads high currents. It should be

wide and short to minimize track impedance to reduce losses and EMI.

VDD pin This pin is designed to provide signal supply voltage to the device. There is no specific

requirement for its related track design.

VLX1/VLX2 pins External coil is connected on those pins. It should be placed as closed as possible to the device

in order minimize resistances which cause looses. These paths lead high currents.

VOUT1 pin3 It is the first output voltage of this device. This path leads high currents. It should be wide and

short to minimize track impedance to reduce losses and EMI.

VOUT2 pin It is the second output voltage of this device. This path leads high currents. It should be wide and

short to minimize track impedance to reduce losses and EMI.

FB1 pin Intended to measure VOUT1 voltage in order to ensure the regulation of this output.

FB2 pin Intended to measure VOUT2 voltage in order to ensure the regulation of this output.

ENABLE pin This is the enable pin of the device. Pulling this pin to ground, forces the device into shutdown

mode. Pulling this pin to VDD enables the device. This pin must be terminated.

MODE/SYNC pin The MODE/SYNC pin is a multipurpose pin which provides mode selection and frequency

synchronization.

The device can also be synchronized to an external clock signal from 600 kHz to 1.5 MHz by the MODE/SYNC pin. During synchronization, the mode is forced to PWM mode and the top switch

turn-on is synchronized to the rising edge of the external clock.

AUTO pin This pin allows the device to automatically switch from PWM to PFM mode following load on both

2 outputs.

STATE pin This output pin informs user in which state the device is working: PWM or PFM mode.

VSEL pin This pin is used to reduce V_{OUT2} (CORE) output voltage in order to reduce the processor power

consumption when entering into sleep mode.

Electrical characteristics STw4141

2 Electrical characteristics

2.1 Absolute maximum ratings

Absolute maximum ratings are those values beyond which damage to the device may occur. Functional operation under these conditions is not implied. All voltages are referenced to GND.

Table 2. Absolute maximum ratings

Symbol	Parameter	Value	Unit
PV_{DD}	Power supply voltage	-0.3 to 6	V
V_{DD}	Signal supply voltage	-0.3 to 6	V
V _{EN}	Enable input	-0.3 to V _{DD}	V
V _{SEL}	Voltage selection	-0.3 to V _{DD}	V
V _{MODE/SYNC}	Operating mode selection/synchronization input	-0.3 to V _{DD}	V
V _{AUTO}	PWM/PFM automatic switch selection	-0.3 to V _{DD}	V
V _{T_MODE}	Test mode selection	-0.3 to V _{DD}	V
V _{STATE}	Operating mode information	-0.3 to V _{OUT1}	V
V _{OUT1} ,FB ₁	Output voltage 1, feedback 1	-0.3 to 3.3	V
V _{OUT2} , FB ₂	Output voltage 2, feedback 2	-0.3 to 3.3	V
VLX ₁	External inductor connection pin 1	-0.3 to V _{DD}	V
VLX ₂	External inductor connection pin 2	-0.3 to 3.3	V
T _A	Operating temperature range	-40 to 85	°C
T _J	Maximum operating junction temperature	150	°C
T _{STG}	Storage temperature range	-65 to 150	°C

2.2 Thermal data

Table 3. Thermal data

Symbol	Parameter	Value	Unit
R _{thJA}	Thermal Resistance Junction-Ambient TFBGA 3x3 mm – 16 bumps - 0.5 mm pitch	150	°C/W

2.3 DC electrical characteristics

Characteristics measured over recommended operating conditions unless otherwise is noted.

All typical values are referred to T_A = 25°C, PVDD = 3.6V, VDD = 3.6V.

Table 4. DC electrical characteristics

Symbol	Parameter	Test Conditions	Min.	Тур	Max.	Unit
PVDD	Power supply voltage		2.7		5.5	V
I _{LIM}	Peak current limit			1.6		Α
V _{OUT1} ⁽¹⁾	Output voltage 1 ⁽²⁾		-3		+3	%
V	Output voltage 2 ⁽²⁾	VSEL = VDD, MODE/SYNC = VDD	-3		+3	%
V _{OUT2}	Output voltage 2 ⁽²⁾	VSEL = GND, MODE/SYNC = VDD	-3		+3	%
I _{OUT1}	Output current 1				200	mA
I _{OUT2}	Output current 2				400	mA
	Quiescent current (PWM)	$\begin{split} &I_{OUT1} = 0 \text{ mA, } I_{OUT2} = 0 \text{ mA} \\ &EN = \text{VDD, VSEL} = \text{VDD} \\ &MODE/SYNC = \text{VDD,} \\ &AUTO = \text{GND} \end{split}$		600		μА
I _q	Quiescent current (PFM)	I _{OUT1} = 0 mA, I _{OUT2} = 0 mA EN = VDD, VSEL = VDD MODE/SYNC = GND, AUTO = GND		90		μA
	Shutdown current	EN = GND, VSEL = GND MODE/SYNC = GND, AUTO = GND		1	5	μА
Enable fu	nctions					
V _{EN} H	Enable threshold high		0.9			V
V _{EN} L	Enable threshold low				0.4	V
Mode/syr	c functions				1	•
V _{M/S} H	MODE/SYNC threshold high		0.9			V
V _{M/S} L	MODE/SYNC threshold low				0.4	V
VSEL fun	ctions	•	·		•	•
V _{SEL} H	Voltage selection threshold high		0.9			V
V _{SEL} L	Voltage selection threshold low				0.4	V

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Electrical characteristics STw4141

Table 4. DC electrical characteristics (continued)

Symbol	Parameter	Test Conditions	Min.	Тур	Max.	Unit		
Auto func	Auto functions							
V _{AUTO} H	Voltage selection threshold high		0.9			V		
V _{AUTO} L	Voltage selection threshold low				0.4	V		
State fund	ctions							
V _{STATE} H	Voltage selection threshold high	$R_{Lmax} = 100k, C_{Lmax} = 10pF$	0.7V _{OUT1}			V		
V _{STATE} L	Voltage selection threshold low	$R_{Lmax} = 100k, C_{Lmax} = 10pF$			0.3 V _{OUT1}	V		

^{1.} $VOUT1 \ge VOUT2$. This condition must always be valid.

2.4 Dynamic electrical characteristics

Characteristics measured over recommended operating conditions unless otherwise is noted.

All typical values are referred to $T_A = 25$ °C, PVDD = 3.6V, VDD = 3.6V.

Table 5. Dynamic electrical characteristics

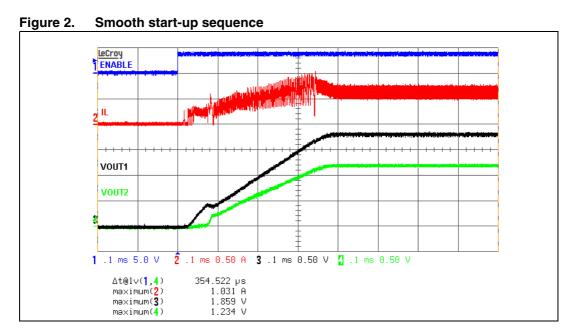
Symbol	Parameter	Test Conditions	Min.	Тур	Max.	Unit
f_{SW}	Switching frequency			900		kHz
f _{SYNC}	Sync mode frequency		600		1500	kHz
Ts	Settling time (soft start)			400		μs
TS2	Settling time V _{OUT2} (reduced)/V _{OUT2}	VSEL change from GND to VDD		80		μs

^{2.} Output voltage accuracy excludes line and load transients

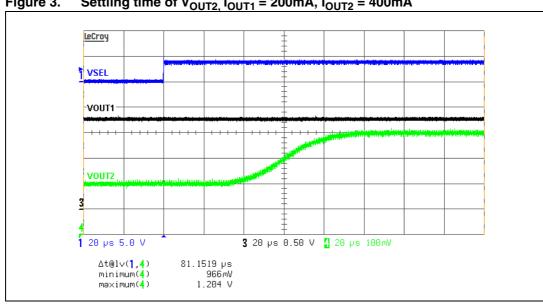
2.5 Soft start

To avoid spikes on battery during STw4141 start-up sequence, a smooth start-up is implemented. Reference voltage grows up less than 600 µs until it achieves is final target. Therefore, STw4141 start up is smooth and secure for the overall mobile phone.

Figure 2 illustrates a smooth start up sequence where V_{IN} = 3.6V, V_{OUT1} = 1.8V@ 200mA, $V_{OUT2} = 1.2V@400mA.$



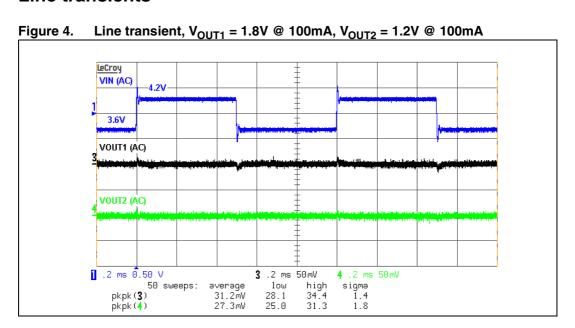
Settling time of VOUT2 2.6



Settling time of V_{OUT2} , $I_{OUT1} = 200$ mA, $I_{OUT2} = 400$ mA Figure 3.

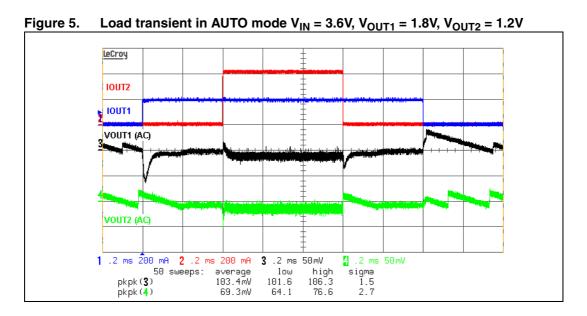
Electrical characteristics STw4141

2.7 Line transients



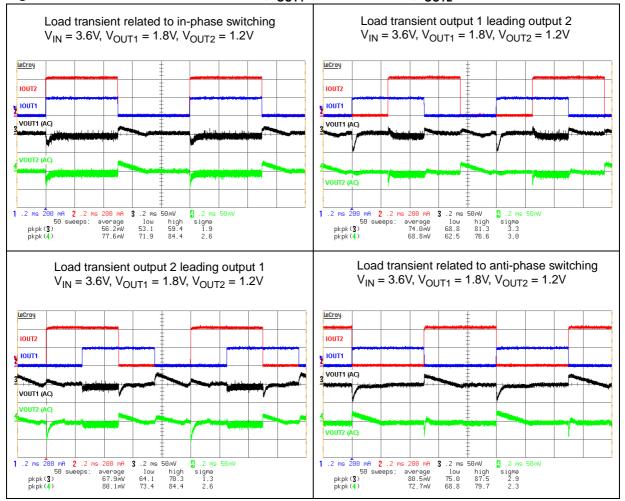
2.8 Load transients in AUTO mode

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2.9 Load transients in PWM mode

Figure 6. Load transient in PWM mode, $I_{OUT1} = 1mA$ to 200mA, $I_{OUT2} = 1mA$ to 400mA



Electrical characteristics STw4141

2.10 Switching between PFM and PWM in FORCED mode



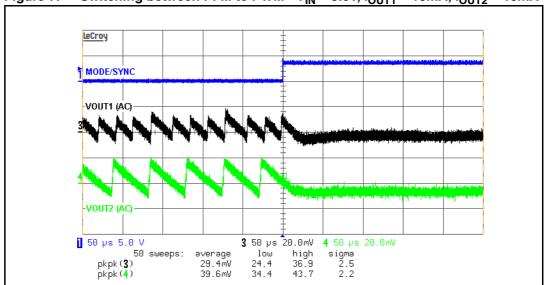
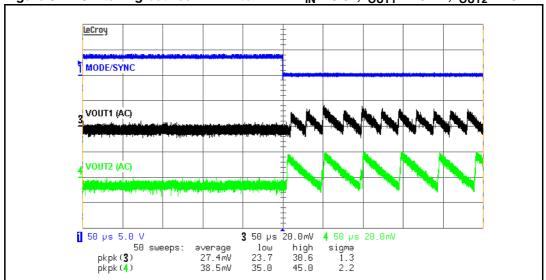


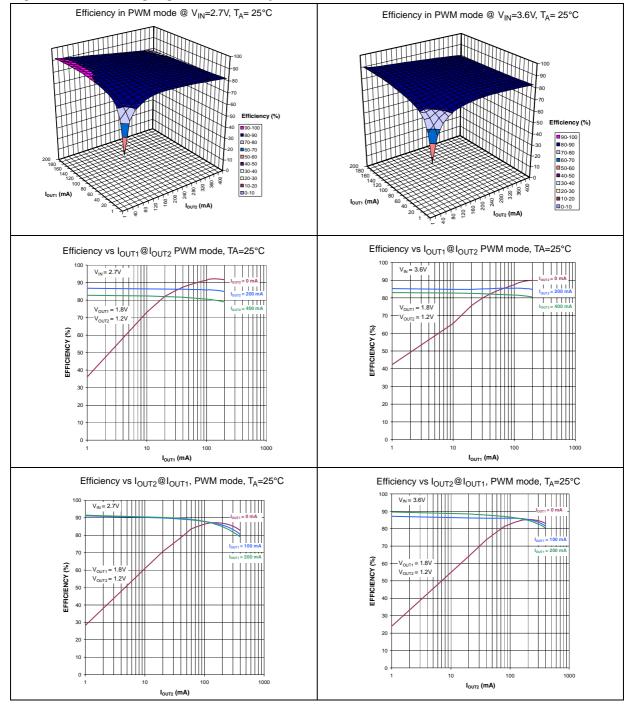
Figure 8. Switching between PWM to PFM - V_{IN} = 3.6V, I_{OUT1} = 10mA, I_{OUT2} = 10mA



2.11 Efficiency in PWM

The efficiency of a switching regulator is equal to the total output power divided by the input. STw4141 has high efficiency up to 92% (for the 2 outputs). Efficiency curve is flat over the output current range.

Figure 9. Switching regulator efficiency in PWM mode



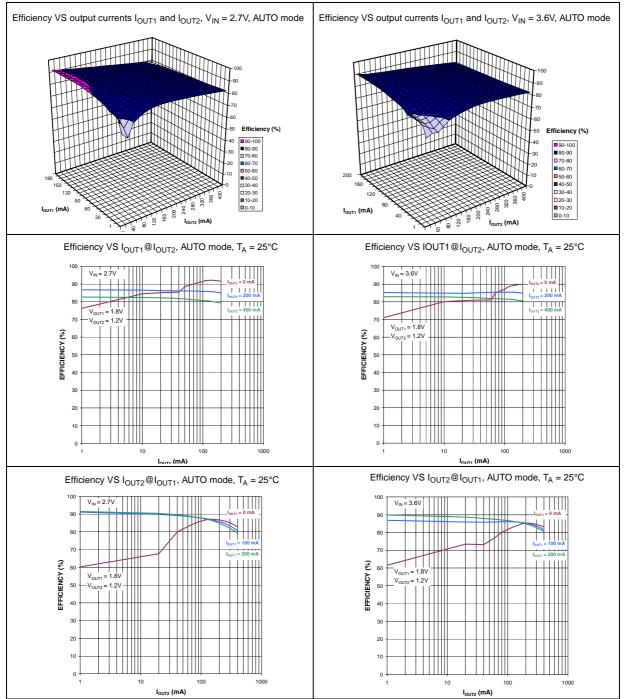


Electrical characteristics STw4141

2.12 Efficiency in AUTO

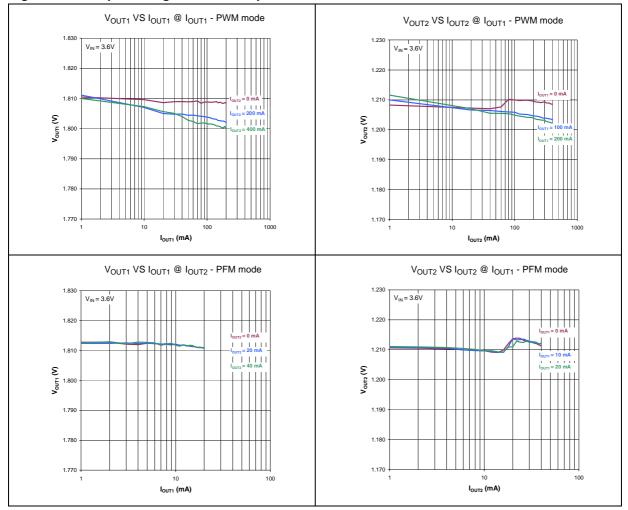
The efficiency of a switching regulator is equal to the total output power divided by the input. STw4141 has high efficiency up to 92% (both outputs) and always higher than 70% for output currents higher than 1mA.

Figure 10. Switching regulator efficiency in auto mode



2.13 Output voltages versus output currents in PWM and PFM

Figure 11. Output voltages versus output currents in PWM and PFM



3 Functional description

Introduction

The STw4141 is an easy to use, single coil dual outputs step down DC/DC converter optimized to supply low-voltage to CPUs or DSPs in cell phones and other miniature devices powered by single cell lithium-ion or 3 cell NiMH/NiCd batteries. It provides two different output voltages with high efficiency operation in a wide range of output currents. The device offers high DC voltage regulation accuracy and load transient response to satisfy demanding processor core supply. The converter is based on voltage mode buck architecture using PWM and PFM operation modes.

At light load currents, the device can operate in PFM mode to maintain high efficiency over the entire load current range. Switching between PWM and PFM modes can be done automatically or can be forced by external pins (AUTO and MODE/SYNC). Externally synchronized or fixed frequency (internal oscillator) PWM mode offers full output current capability while minimizing interference to sensitive RF and data acquisition circuits.

PWM and PFM mode operation

PWM (Pulse Width Modulation) mode is intended for normal load to high load currents. Energy is delivered to the load with an accurate and defined frequency of 900 kHz. PFM (Pulse Frequency Modulation) mode is intended for low load currents to maintain high efficiency conversion.

Forced mode: When AUTO pin is LOW, the operating mode is selectable by the user itself. It means that system controls the behavior of the STw4141 according to processor needs. STw4141 is switched from PWM to PFM mode using MODE/SYNC pin (refer to *Section 4.1*). **Automatic PWM / PFM switch:** When AUTO pin is HIGH, the operating mode is directly controlled by internal digital circuit according to processor needs. The device switches from PWM to PFM by itself if sum of output currents is lower than approximately 100 mA during at least 16 clock cycles. The device can be forced to PWM mode connecting MODE/SYNC to HIGH level. (see *Section 2.8* and *Section 2.9*).

Current limiter

This protection limits the current flowing through coil. As soon as ILIM is detected, the duty cycle is terminated and prevents the coil current against rising above peak current limit. There is no reset of device.

Short circuit protection

It protects the device against short-circuit at output terminals. When one or both output voltages are decreased by 0.7 V below their nominal output values the device enters into reset followed by soft start sequence.

Thermal shutdown protection

Thermal shutdown protects the device against damage due to overheating when maximum operating junction temperature is exceeded. The device is kept in reset until junction temperature decreases by 25°C approximately.

4 Application information

4.1 User mode details

The following table describes the different user modes available. Depending on the application constraints (processor I/O pins available) and expected efficiency, PWM or PFM mode are forced or automatically controlled by STw4141 internal digital gates.

Table 6. STw4141 available user modes

Mode	User mode/pins	EN	AUTO	MODE/SYNC
OFF	Shutdown	L	X	Х
	Forced PFM	Н	L	L
FORCED	Forced PWM	Н	L	Н
	Forced PWM and synchronized external clock	Н	L	CLK
	Auto mode	Н	Н	L
AUTO	Forced PWM	Н	Н	Н
	Forced PWM and synchronized external clock	Н	Н	CLK

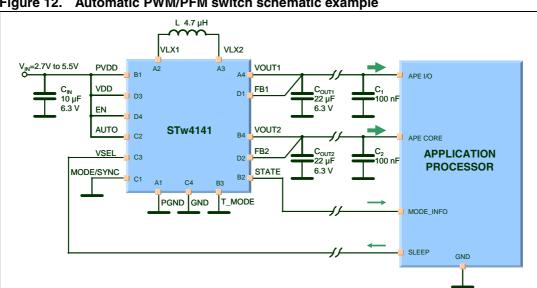
Table 7. Operating mode information (STATE pin - digital output)

Operation mode	State pin voltage level
PFM	VOUT1
PWM	GND

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4.2 **Automatic PWM/PFM mode**

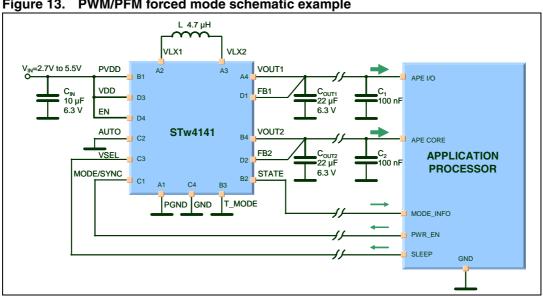
This user mode is designed to allow STw4141 to switch automatically between PWM and PFM modes. This feature improves the application efficiency because STw4141 enters in PFM mode according to application processor current consumption.



Automatic PWM/PFM switch schematic example Figure 12.

4.3 User selected PWM/PFM mode

STw4141 PWM/PFM mode can also be controlled by the application processor. This feature is accessible through MODE/SYNC pin state. It is useful to users who want to use STw4141 with the modern digital processor. Therefore, MODE/SYNC pin is connected to SLEEP mobile phone signal.



PWM/PFM forced mode schematic example

4.4 External clock synchronization

L 4.7 µH VLX1 VLX2 V_{IN}=2.7V to 5.5V VOUT1 APE I/O C_{IN} 10 μF VDD FB1 D3 100 nF ΕN AUTO STw4141 VOUT2 C2 FB2 **APPLICATION PROCESSOR** MODE/SYNC STATE 63 V C1 PGND GND T_MODE CLK SLEEF GND

Figure 14. Application using external clock synchronization

4.5 Checking transient response versus external components

The regulator loop response can be checked by looking at the load transient response. Switching regulators take several cycles to respond to a step in load current. When a load step occurs, V_{OUT} is immediately shifted by an amount equal to I_{LOAD} x ESR, where ESR is the equivalent series resistance of C_{OUT} . I_{LOAD} also begins to charge or discharge C_{OUT} generating a feedback error signal used by the regulator to return V_{OUT} to its steady-state value. In order to improve the transient response, it is better to use two 10 μF ceramic capacitors on each output to reduce ESR.

4.6 Bill of Material

4.6.1 Inductor selection

The choice of which inductor to use depends on the price and size versus performance required with the STw4141 application. *Table 7* shows some typical surface mount inductors that work well in STw4141 applications.

Part number	Supplier	Value (µH)	R _{DC} (Ω)	Max DC current (mA)	Size (mm) W x L x H
VFL4012A-4R7M1R1	TDK	4.7	0.14	1100	3.5 x 3.7 x 1.2
VFL3012A-4R7MR74	TDK	4.7	0.16	740	2.6 x 2.8 x 1.2
744031004	WUERTH	4.7	0.085	900	3.8 x 3.8 x 1.8

Table 8. Bill of material: inductor selection

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4.6.2 Input capacitor (CIN selection)

Input capacitor of 10 μ F ceramic low ESR capacitor should be used to reduce switching losses. It should be placed as close as possible to supply pins VDD and PVDD. The connection traces should be wide and short to minimize impedance.

4.6.3 Output capacitors (COUT selection)

The selection of COUT is driven by the required ESR to minimize voltage ripple and load step transients. There are two possibilities for output capacitors: either a 22 μF is connected to ground or two 10 μF ceramic are used to reduce ESR and switching losses. The capacitor should be placed as close as possible to V_{OUT_X} pins. The connection traces should be wide and short to minimize impedance.

4.6.4 Capacitors selection

Table 9. Bill of Material: capacitor selection

Component	Supplier	Part number	Value	Case size
	MURATA	GRM188R60J106ME47D	10 μF, 6.3V	0603
		GRM21BR60J106ME15L	10 μF, 6.3V	0805
C _{IN}	TAIYO YUDEN	JMK212BJ106MG-T	10 μF, 6.3V	0805
	TDK	C1608X5R0J106MT	10 μF, 6.3V	0603
		C2012X5R0J106MT	10 μF, 6.3V	0805
	MURATA	GRM188R60J106ME47D	2 x 10 μF, 6.3V	2 x 0603
		GRM21BR60J106ME15L	2 x 10 μF, 6.3V	2 x 0805
C _{OUT1} , C _{OUT2}		GRM21BR60J226ME15L	22 μF, 6.3V	0805
	TAYIO YUDEN	JMK212BJ106MG-T	2 x 10 μF, 6.3V	2 x 0805
		JMK212BJ226MG-T	22 μF, 6.3V	0805
	TDK	C1608X5R0J106MT	2 x 10 μF, 6.3V	2 x 0603
		C2012X5R0J106MT	2 x 10 μF, 6.3V	2 x 0805
		C2012X5R0J226MT	22 μF, 6.3V	0805

4.7 PCB layout considerations

The Printed Circuit Board layout must include the following consideration: Current paths carrying high currents (bold lines in *Figure 15*) must be wide and short to minimize impedance in order to reduce looses and EMI.

Small currents flow through voltage paths. No specific care is requested about voltage paths but it is recommended to follow the general rules for PCB routing to reduce influence of external and internal interferences.

L 4.7 µH HIGH CURRENT PATH VLX1 VLX2 V_{OUT1}=1.8V V_{IN} =2.7V to 5.5V **PVDD** VOUT1 АЗ B1 C_{IN} 10 µF VDD FB1 C_{ΟUT1} 22 μF 6.3 V D1 D3 6.3 V ΕN V_{OUT2}=1.0V/1.2 AUTO STw4141 VOUT2 C_{OUT2} 22 μF VSEL FB2 D2 MODE/SYNC 6.3 V STATE B2 C1 В3

PGND GND

T_MODE

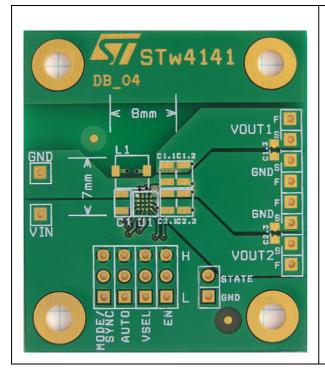
Figure 15. Board layout track length and width

4.7.1 PCB layout

Figure 16 and *Figure 17* show the PCB layout. All components are on the top side of the board.

Figure 16. Demoboard top layer

Figure 17. Demoboard assembled with 22 μF output capacitor





4.7.2 TFBGA16 internal bumps access

Pad centers are at 500 μ m distance. Pad diameter is 275 μ m. The distance between two adjacent pad edges is only 225 μ m. We recommend a distance for lead-out signals from the center of pad matrix by 75 μ m wide trace. Isolation distance in this case is 75 μ m (see *Figure 18* and *Figure 19*).

Figure 18. TFBGA16 ball pad spacing and track parameters to internal pads

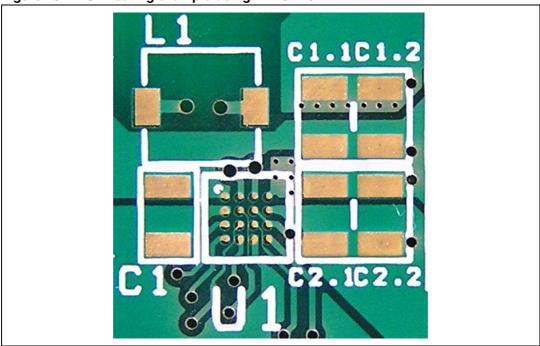


Figure 19. PCB routing example using TFBGA16

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5 Package outline and mechanical data

Table 10. TFBGA 3x3x1.20 16 F4x4 0.50. Package code: L0 - JEDEC/EAIJ

Ref	Min.	Тур.	Max.
Α	1.01		1.20 (<i>Note 1</i>)
A1	0.15		
A2		0.82	
b	0.25	0.30	0.35
D	2.85	3.00	3.15
D1		1.50	
Е	2.85	3.00	3.15
E1		1.50	
е		0.50	
ddd	0.85		0.08
eee			0.15
fff			0.05

Note: 1 Max mounted height is 1.12 mm. Based on a 0.28 mm ball pad diameter. Solder paste is 0.15 mm thickness and 0.28 mm diameter.

2 TFBGA stands for **Thin Profile Fine Pitch Ball Grid Array**.

Thin profile: The total profile height (DIm A) is measured from the seating plane to the top of the component.

A = 1.01 to 1.20 mm

Fine pitch < 1.00 mm pitch.

3 The tolerance of position that controls the location of the pattern of balls with respect to datums A and B.

For each ball there is a cylindral tolerance zone eee perpendicular to datum C and located on true position with respect to datums A and B as defined by e. The axis perpendicular to datum C of each ball must lie within this tolerance zone.

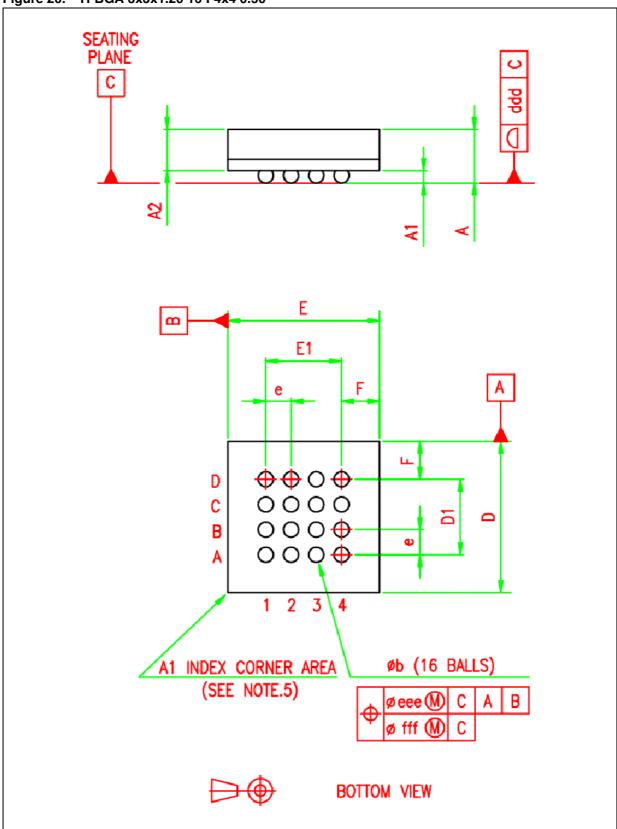
4 The tolerance of position that controls the location of the balls within the matrix with respect to each other. For each ball there is a cylindrical tolerance zone fff perpendicular to datum C and located on true position as defined by e. The axis perpendicular to datum C of each ball must lie within the tolerance zone.

Each tolerance zone fff in the array is contained entirely in the respective above eee zone above

The axis of each ball must be simultaneously in both tolerance zones.

5 Leadfree package according to JEDEC JESD-020-C

Figure 20. TFBGA 3x3x1.20 16 F4x4 0.50



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Ordering information STw4141

6 Ordering information

Table 11. STw4141 order codes

	Package	Package marking	Output voltage options ⁽¹⁾			
Part number			V _{оυт1} ⁽²⁾ (I/O)	V _{OUT2} (CORE)	V _{OUT2reduced} (CORE)	Packing
STw41411		STA1	1.8 V	1.2 V	1.0 V	Tray
STw41411/T		STA1	1.8 V	1.2 V	1.0 V	Tape and reel
STw41412		STA2	1.8 V	1.3 V	1.0 V	Tray
STw41412/T		STA2	1.8 V	1.3 V	1.0 V	Tape and reel
STw41413	TFBGA	STA3	1.8 V	1.4 V	1.2 V	Tray
STw41413/T		STA3	1.8 V	1.4 V	1.2 V	Tape and reel
STw41414		STA4	1.8 V	1.5 V	1.3 V	Tray
STw41414/T	3x3x1.2 16 balls	STA4	1.8 V	1.5 V	1.3 V	Tape and reel
STw41415		STA5	1.8 V	1.25 V	1.0 V	Tray
STw41415/T		STA5	1.8 V	1.25 V	1.0 V	Tape and reel
STw41416		STA6	1.8 V	1.35 V	1.0 V	Tray
STw41416/T		STA6	1.8 V	1.35 V	1.0 V	Tape and reel
STw41417		STA7	1.5 V	1.3 V	1.0 V	Tray
STw41417/T		STA7	1.5 V	1.3 V	1.0 V	Tape and reel

^{1.} The output configuration which will be introduced in production will be only those related to customer design-in.

^{2.} VOUT1≥VOUT2 THIS CONDITION MUST BE ALWAYS VALID.

STw4141 Revision history

7 Revision history

Table 12. Document revision history

Date	Revision	Changes
24-Jan-2006	1	Initial release.
9-Jun-2006	2	Updates in <i>Table 8: Bill of material: inductor selection</i> Updates in <i>Table 9: Bill of Material: capacitor selection</i> Moved <i>Ordering information</i> to the end of the document and updated the order codes

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