



## SINGLE-SUPPLY DUAL OPERATIONAL AMPLIFIER

### ■ GENERAL DESCRIPTION

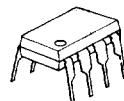
The NJM2904 consists of two independent, high gain, internally frequency compensated operation amplifiers which were designed specifically to operate from a single power supply over a wide range of voltages. Operation from split power supplies is also possible and the low power supply current drain is independent of the magnitude of the power supply voltage.

Application areas include transducer amplifiers, DC gain blocks, and all the conventional op amp circuits which now can be more easily implemented in single power supply systems. For example, the NJM2904 can be directly operated off of the standard +5V power supply voltage which is used in digital systems and will easily provide the required interface electronics without requiring the additional  $\pm 15V$  power supplies.

### ■ FEATURES

- Single Supply
- Operating Voltage                (+3V ~ +32V)
- Low Operating Current        (0.7mA typ.)
- Slew Rate                        (0.5V/ $\mu$ s typ.)
- Bipolar Technology
- Package Outline                DIP8, DMP8, SIP8, SSOP8

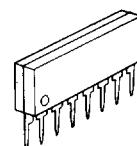
### ■ PACKAGE OUTLINE



NJM2904D



NJM2904M

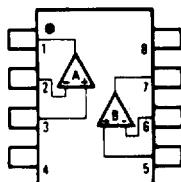


NJM2904L



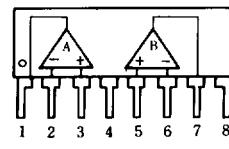
NJM2904V

### ■ PIN CONFIGURATION



#### PIN FUNCTION

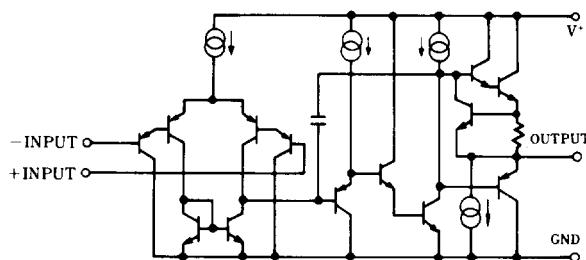
1. A OUTPUT
2. A-INPUT
3. A+INPUT
4. GND
5. B+INPUT
6. B-INPUT
7. B OUTPUT
8. V+



NJM2904L

NJM2904D  
NJM2904M  
NJM2904V

### ■ EQUIVALENT CIRCUIT (1/2 Shown)





## ■ ABSOLUTE MAXIMUM RATINGS

(Ta=25°C)

PARAMETER	SYMBOL	RATINGS	UNIT
Supply Voltage	V <sup>+</sup> (V <sup>+/V<sup>-</sup></sup> )	32(or ±16)	V
Differential Input Voltage	V <sub>ID</sub>	32	V
Input Voltage	V <sub>I</sub>	-0.3~+32	V
	P <sub>D</sub>	(DIP8) 500 (DMP8) 300 (SSOP8) 300 (SIP8) 800	mW
Operating Temperature Range	T <sub>opr</sub>	-40~+85	°C
Storage Temperature Range	T <sub>stg</sub>	-50~+120	°C

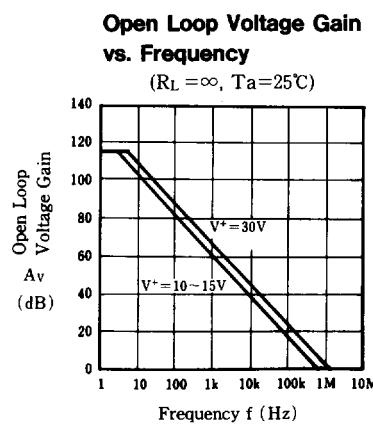
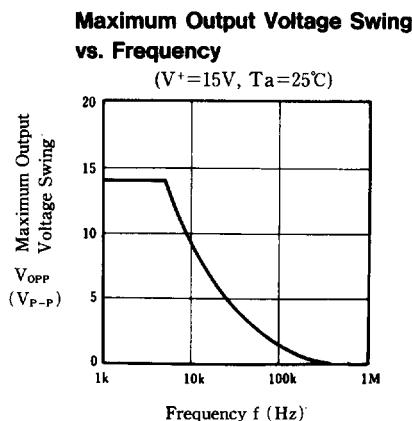
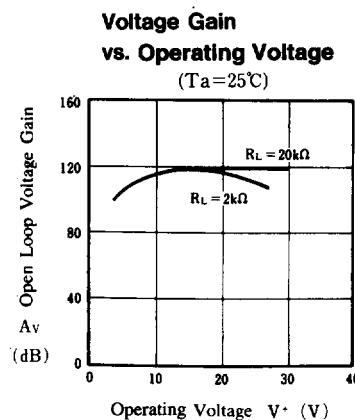
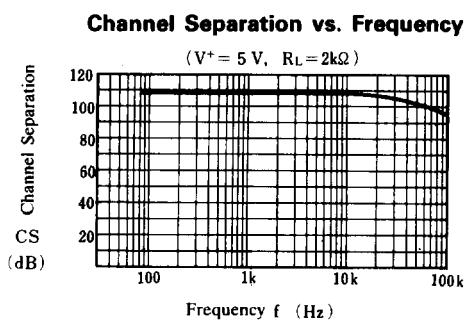
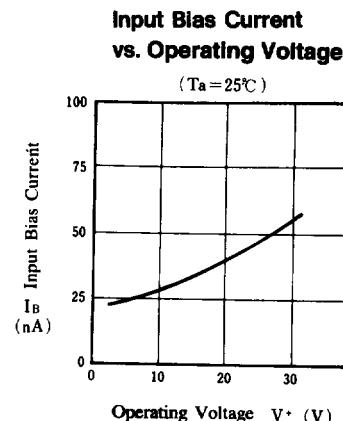
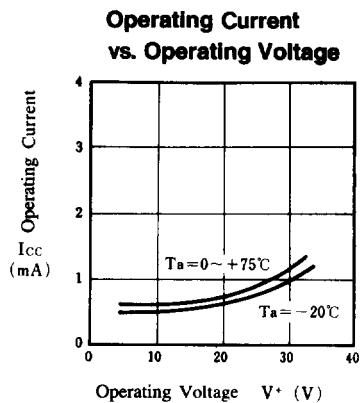
## ■ ELECTRICAL CHARACTERISTICS

(Ta=25°C V<sup>+</sup>=5V)

PARAMETER	SYMBOL	TEST CONDITION	MIN.	TYP.	MAX.	UNIT
Input Offset Voltage	V <sub>IO</sub>	R <sub>S</sub> =0Ω	—	2	7	mV
Input Offset Current	I <sub>IO</sub>		—	5	50	nA
Input Bias Current	I <sub>B</sub>		—	25	250	nA
Large Signal Voltage Gain	A <sub>V</sub>	R <sub>L</sub> ≥2kΩ	—	100	—	dB
Maximum Output Voltage Swing	V <sub>OPP</sub>	R <sub>L</sub> =2kΩ	3.5	—	—	V
Input Common Mode Voltage Range	V <sub>ICM</sub>		0~3.5	—	—	V
Common Mode Rejection Ratio	CMR		—	85	—	dB
Supply Voltage Rejection Ratio	SVR		—	100	—	dB
Output Source Current	I <sub>SOURCE</sub>	V <sub>IN</sub> <sup>+</sup> =1V, V <sub>IN</sub> <sup>-</sup> =0V	20	30	—	mA
Output Sink Current	I <sub>SINK</sub>	V <sub>IN</sub> <sup>+</sup> =0V, V <sub>IN</sub> <sup>-</sup> =1V	8	20	—	mA
Channel Separation	CS	f=1k~20kHz, Input Referred	—	120	—	dB
Operating Current	I <sub>cc</sub>	R <sub>L</sub> =∞	—	0.7	1.2	mA
Slew Rate	SR		—	0.5	—	V/μs
Unity Gain Bandwidth	f <sub>t</sub>		—	0.2	—	MHz

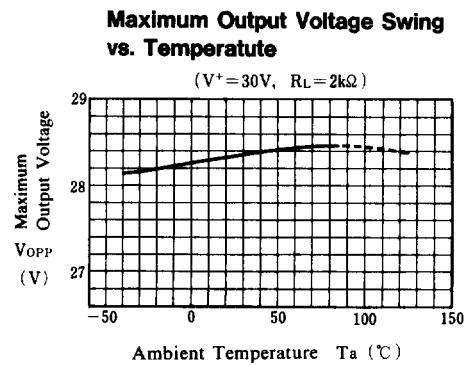


## ■ TYPICAL CHARACTERISTICS

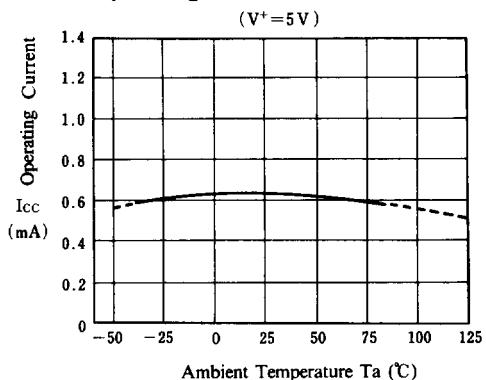




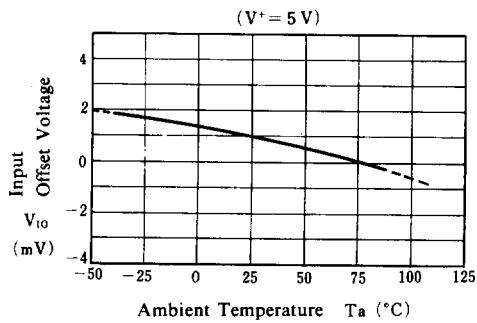
## ■ TYPICAL CHARACTERISTICS



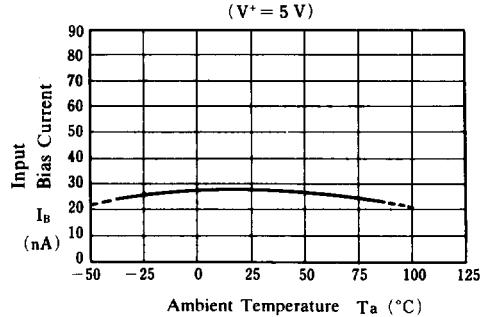
### Operating Current vs. Temperature



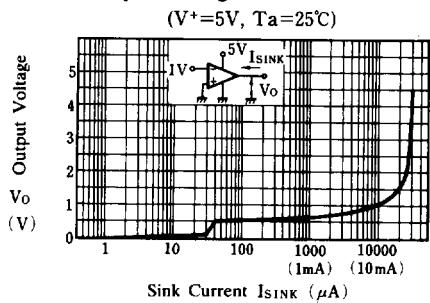
### Input Offset Voltage vs. Temperature



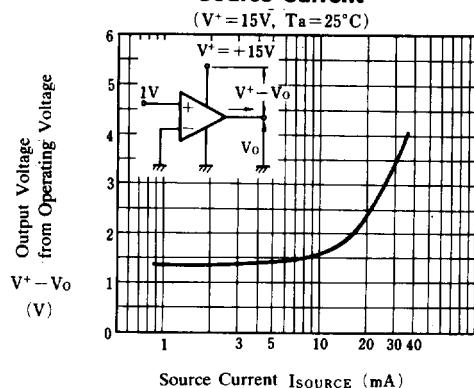
### Input Bias Current vs. Temperature



### Output Voltage vs. Sink Current

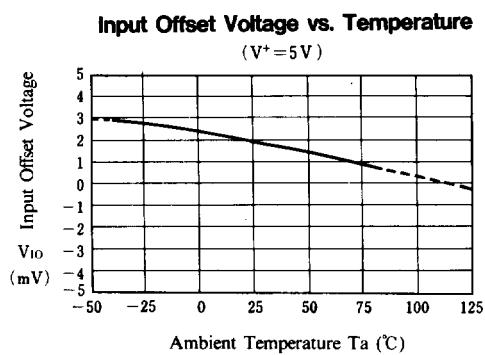
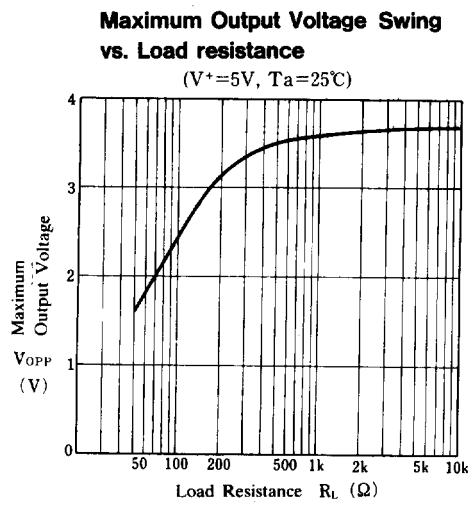
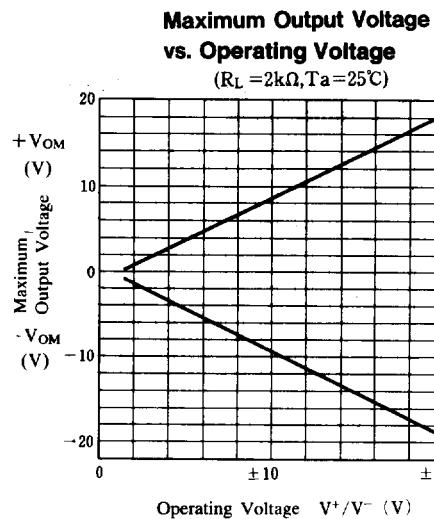
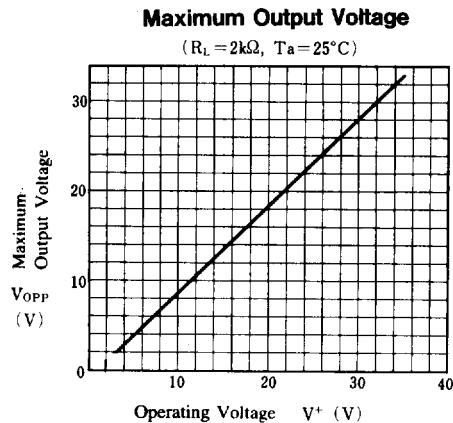


### Source Current





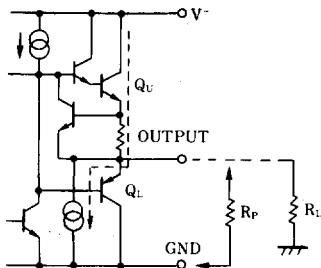
■ TYPICAL CHARACTERISTICS





## ■ APPLICATION

- Improvement of Cross-over Distortion  
Equivalent circuit at the output stage



NJM2904, in its static state (No in and output condition) when design,  $Q_U$  being biassed by constant current (breakdown beam) yet,  $Q_L$  stays OFF.

While using with both power source mode, the cross-over distortion might occur instantly when  $Q_L$  ON.

There might be cases when application for amplifier of audio signals, not only distortion but also the apparent frequency bandwidth being narrowed remarkably.

It is advisable especially when using both power source mode, constantly to use with higher current on  $Q_U$  than the load current (including feedback current), and then connect the pull-down resistor  $R_P$  at the part between output and GND pins.

