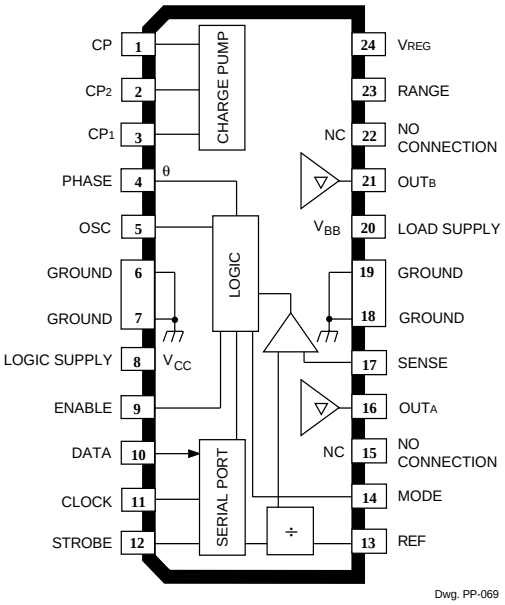
3958

Data Sheet
29319.31

PRELIMINARY INFORMATION
(subject to change without notice)
May 24, 1999

DMOS FULL-BRIDGE PWM MOTOR DRIVER

A3958SLB



Note that the A3958SLB(SOIC) and A3958SB (DIP) do not share a common terminal assignment.

ABSOLUTE MAXIMUM RATINGS

Load Supply Voltage, V_{BB} ..... **50 V**
Output Current, I_{OUT} ..... **±2.0 A**
Logic Supply Voltage, V_{DD} ..... **7.0 V**
Input Voltage, V_{IN} .... **-0.3 V to V_{DD} + 0.3 V**
Sense Voltage, V_S ..... **0.5 V**
Reference Voltage, V_{REF} ..... **2.7 V**
Package Power Dissipation (T_A = 25°C), P_D
A3958SB ..... **3.1 W***
A3958SLB ..... **2.2 W***
Operating Temperature Range,
T_A ..... **-20°C to +85°C**
Junction Temperature,
T_J ..... **+150°C**
Storage Temperature Range,
T_S ..... **-55°C to +150°C**

Output current rating may be limited by duty cycle, ambient temperature, and heat sinking. Under any set of conditions, do not exceed the specified current rating or a junction temperature of 150°C.

* Per SEMI G42-88 Specification.

Designed for pulse-width modulated (PWM) current control of dc motors, the A3958SB and A3958SLB are capable of continuous output currents to ±2 A and operating voltages to 50 V. Internal fixed off-time PWM current-control timing circuitry can be programmed via a serial interface to operate in slow, fast, and mixed current-decay modes.

PHASE and ENABLE input terminals are provided for use in controlling the speed and direction of a dc motor with externally applied PWM-control signals. The ENABLE input can be programmed via the serial port to PWM the bridge in fast or slow current decay. Internal synchronous rectification control circuitry is provided to reduce power dissipation during PWM operation.

Internal circuit protection includes thermal shutdown with hysteresis, and crossover-current protection. Special power-up sequencing is not required.

The A3958SB/SLB is supplied in a choice of two power packages, a 24-pin plastic DIP with a copper batwing tab (package suffix 'B'), and a 24-lead plastic SOIC with a copper batwing tab (package suffix 'LB'). In both cases, the power tab is at ground potential and needs no electrical isolation.

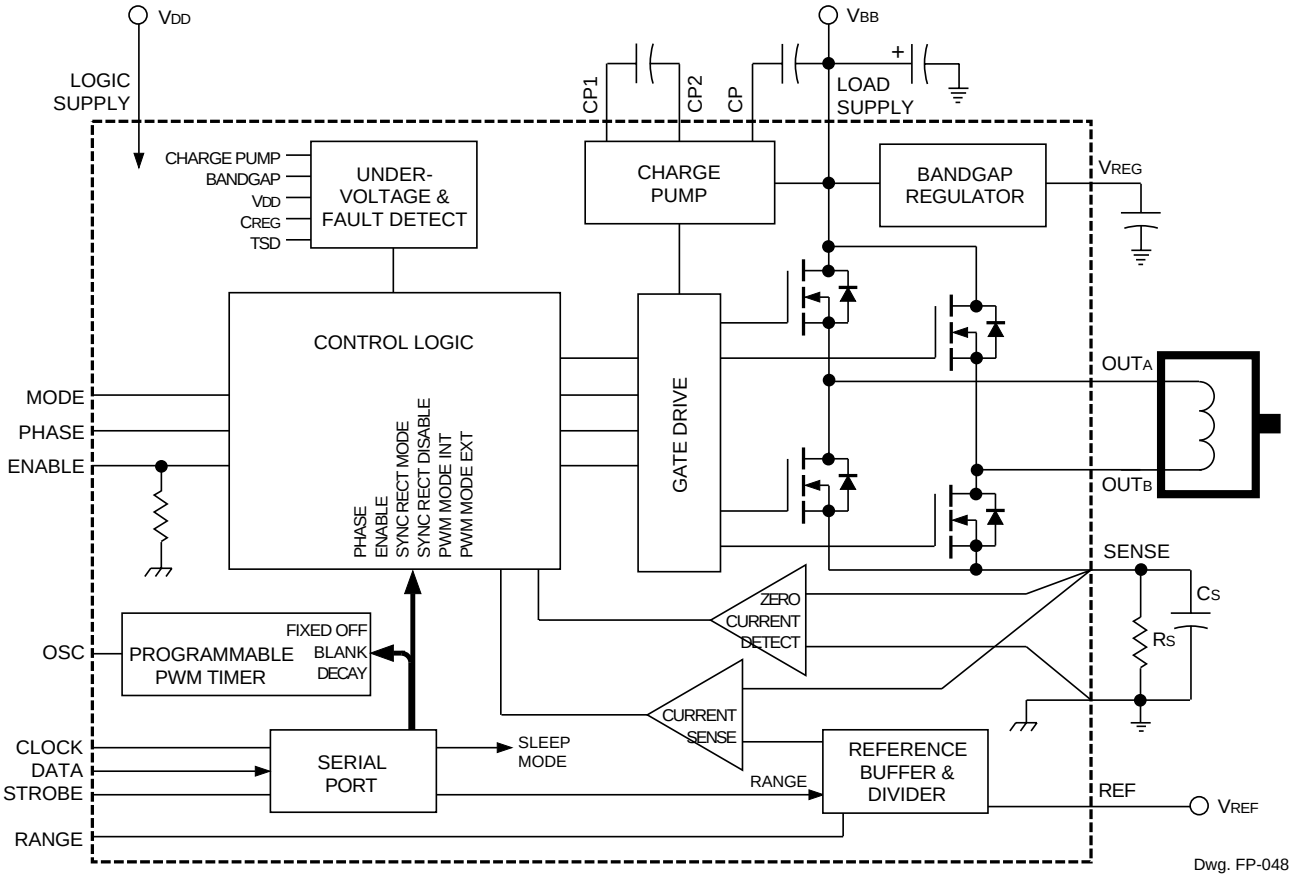
FEATURES

- ±2 A, 50 V Continuous Output Rating
- Low $r_{DS(on)}$ Outputs (270 mΩ, Typical)
- Programmable Mixed, Fast, and Slow Current-Decay Modes
- Serial Interface Controls Chip Functions
- Synchronous Rectification for Low Power Dissipation
- Internal UVLO and Thermal-Shutdown Circuitry
- Crossover-Current Protection

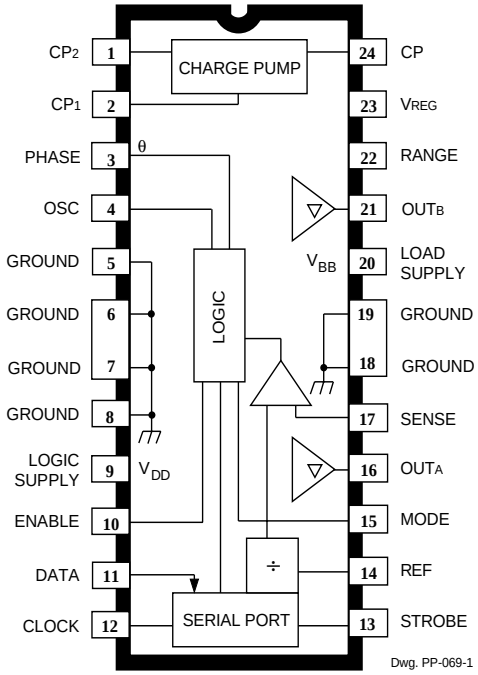
Always order by complete part number:

Part Number	Package	$R_{\theta JA}$	$R_{\theta JT}$
A3958SB	24-pin batwing DIP	40°C/W	6°C/W
A3966SLB	24-lead batwing SOIC	56°C/W	6°C/W

FUNCTIONAL BLOCK DIAGRAM



Dwg. FP-048



Dwg. PP-069-1

A3958SB

Note that the A3958SLB (SOIC) and A3958SB (DIP) do not share a common terminal assignment.

Allegro MicroSystems, Inc.
115 Northeast Cutoff, Box 15036
Worcester, Massachusetts 01615-0036 (508) 853-5000

ELECTRICAL CHARACTERISTICS at T_A = +25°C, V_{BB} = 50 V, V_{DD} = 5.0 V, V_{SENSE} = 0.5 V, f_{PWM} < 50 kHz (unless noted otherwise)

Characteristics	Symbol	Test Conditions	Limits			
			Min.	Typ.	Max.	Units
Output Drivers						
Load Supply Voltage Range	V_{BB}	Operating, I_{OUT} = ±2 A, L = 3 mH	20	–	50	V
		During sleep mode	0	–	50	V
Output Leakage Current	I_{DSS}	V_{OUT} = V_{BB}	–	<1.0	20	µA
		V_{OUT} = 0 V	–	<-1.0	-20	µA
Output On Resistance	$r_{DS(on)}$	Source driver, I_{OUT} = -2 A	–	270	300	mΩ
		Sink driver, I_{OUT} = 2 A	–	270	300	mΩ
Body Diode Forward Voltage	V_F	Source diode, I_F = -2 A	–	1.2	1.6	V
		Sink diode, I_F = 2 A	–	1.2	1.6	V
Load Supply Current	I_{BB}	f_{PWM} < 50 kHz	–	4.0	7.0	mA
		Charge pump on, outputs disabled	–	2.0	5.0	mA
		Sleep Mode	–	–	20	µA
Control Logic						
Logic Supply Voltage Range	V_{DD}	Operating	4.5	5.0	5.5	V
Logic Input Voltage	$V_{IN(1)}$		2.0	–	–	V
	$V_{IN(0)}$		–	–	0.8	V
Logic Input Current (all inputs except ENABLE)	$I_{IN(1)}$	V_{IN} = 2.0 V	–	<1.0	20	µA
	$I_{IN(0)}$	V_{IN} = 0.8 V	–	<-2.0	-20	µA
ENABLE Input Current	$I_{IN(1)}$	V_{IN} = 2.0 V	–	40	100	µA
	$I_{IN(0)}$	V_{IN} = 0.8 V	–	–	0	µA
OSC input frequency	f_{OSC}	Operating	2.9	–	6.1	MHz
OSC input duty cycle	dc_{OSC}	Operating	40	–	60	%
OSC input hysteresis	–	Operating	200	–	400	mV
Input Hysterisis	–	All digital inputs except OSC	50	–	100	mV
Reference Input Volt. Range	V_{REF}	Operating	0.0	–	2.6	V
Reference Input Current	I_{REF}	V_{REF} = 2.5 V	–	–	±0.5	µA
Comparator Input Offset Volt.	V_{IO}	V_{REF} = 0 V	–	0	±5.0	mV

Continued next page …

ELECTRICAL CHARACTERISTICS at T_A = +25°C, V_{BB} = 50 V, V_{DD} = 5.0 V, V_{SENSE} = 0.5 V, f_{PWM} < 50 kHz (unless noted otherwise), continued.

Characteristics	Symbol	Test Conditions	Limits Min.	Typ.	Max.	Units
Control Logic						
Buffer Input Offset Volt.	V_{IO}		–	0	±15	mV
Reference Divider Ratio	–	D14 = High	9.9	10	10.2	–
		D14 = Low	4.95	5.0	5.05	–
Propagation Delay Times	t_{pd}	I_{OUT} = ±2.0 A, 50% to 90%:				
		PWM change to source ON	–	600	–	ns
		PWM change to source OFF	–	100	–	ns
		PWM change to sink ON	–	600	–	ns
		PWM change to sink OFF	–	100	–	ns
		Phase change to sink ON	–	600	–	ns
		Phase change to sink OFF	–	100	–	ns
		Phase change to source ON	–	600	–	ns
		Phase change to source OFF	–	100	–	ns
Thermal Shutdown Temp.	T_J		–	165	–	°C
Thermal Shutdown Hysteresis	ΔT_J		–	15	–	°C
UVLO Enable Threshold	UVLO	Increasing V_{DD}	3.90	4.2	4.45	V
UVLO Hysteresis	ΔUVLO		0.05	0.10	–	V
Logic Supply Current	I_{DD}	f_{PWM} < 50 kHz	–	6.0	10	mA
		Sleep Mode, Inputs < 0.5 V	–	–	2.0	mA

NOTES: 1. Typical Data is for design information only.
2. Negative current is defined as coming out of (sourcing) the specified device terminal.

FUNCTIONAL DESCRIPTION

Serial Interface. The A3958 is controlled via a 3-wire serial port. The programmable functions allow maximum flexibility in configuring the PWM to the motor drive requirements. The serial port is defined in the following table and text descriptions:

Bit	Function
D0	Blank Time LSB
D1	Blank Time MSB
D2	Off Time LSB
D3	Off Time Bit 1
D4	Off Time Bit 2
D5	Off Time Bit 3
D6	Off Time MSB
D7	Fast Decay Time Bit LSB
D8	Fast Decay Time Bit 1
D9	Fast Decay Time Bit 2
D10	Fast Decay Time MSB
D11	Sync. Rect. Mode
D12	Sync. Rect. Enable
D13	External PWM Mode
D14	Enable
D15	Phase
D16	Reference Range Select
D17	Internal PWM Mode
D18	Test Use Only
D19	Sleep Mode

D0 – D1 Blank Time. The current-sense comparator is blanked when any output driver is switched on, according to the table below. f_{osc} is the oscillator input frequency.

D1	D0	Blank Time
0	0	$4/f_{osc}$
0	1	$6/f_{osc}$
1	0	$12/f_{osc}$
1	1	$24/f_{osc}$

D2 – D6 Fixed-Off Time. A five-bit word sets the fixed-off time for internal PWM current control. The off time is defined by the following equation:

$$t_{off} = (8[1 + N]/f_{osc}) - 1/f_{osc}$$

where N = 0 … 31

For example, with an oscillator frequency of 4 MHz, the off time will be adjustable from 2 µs to 64 µs in increments of 2 µs.

D7 – D10 Fast Decay Time. A four-bit word sets the fast decay portion of the fixed-off time for the internal PWM control circuitry. This will only have impact if the mixed-decay mode is selected (via bit D17 and the MODE input terminal). For $t_{fd} > t_{off}$, the device will effectively operate in the fast-decay mode. The fast decay portion is defined by:

$$t_{fd} = (8[1 + N]/f_{osc}) - 1/f_{osc}$$

where N = 0 … 15

For example, with an oscillator frequency of 4 MHz, the fast decay time will be adjustable from 2 µs to 32 µs in increments of 2 µs.

D11 Synchronous Rectification Mode. The active mode prevents reversal of load current by turning off synchronous rectification when a zero current level is detected. The passive mode will allow reversal of current but will turn off the synchronous rectifier circuit if the load current inversion ramps up to the current limit set by V_{REF}/R_S.

D11	Mode
0	Active
1	Passive

D12 Synchronous Rectification Enable.

D12	Synchronous Rect.
0	Disabled
1	Enabled

D13 External PWM Decay Mode. Bit D13 determines the current-decay mode when using ENABLE chopping for external PWM current control.

D13	Mode
0	Fast
1	Slow

D14 Enable Logic. Bit D14, in conjunction with ENABLE, determines if the output drivers are in the chopped (OFF)(ENABLE = D14) or ON (ENABLE ≠ D14) state.

ENABLE	D14	Mode
0	0	Chopped
1	0	On
0	1	On
1	1	Chopped

FUNCTIONAL DESCRIPTION (continued)

D15 Phase Logic. Bit D15, in conjunction with PHASE, determines if the device is operating in the forward (PHASE ≠ D15) or reverse (PHASE = D15) state.

PHASE	D15	State	OUT_A	OUT_B
0	0	Reverse	Low	High
1	0	Forward	High	Low
0	1	Forward	High	Low
1	1	Reverse	Low	High

D16 G_m Range Select. Bit D16, in conjunction with RANGE, determines if V_{REF} is divided by 5 (RANGE ≠ D16) or by 10 (RANGE = D16).

RANGE	D16	Divider
0	0	÷10
1	0	÷5
0	1	÷5
1	1	÷10

D17 Internal PWM Mode. Bit D17, in conjunction with MODE, selects slow (MODE ≠ D17) or mixed (MODE = D17) current decay.

MODE	D17	Current-Decay Mode
0	0	Mixed
1	0	Slow
0	1	Slow
1	1	Mixed

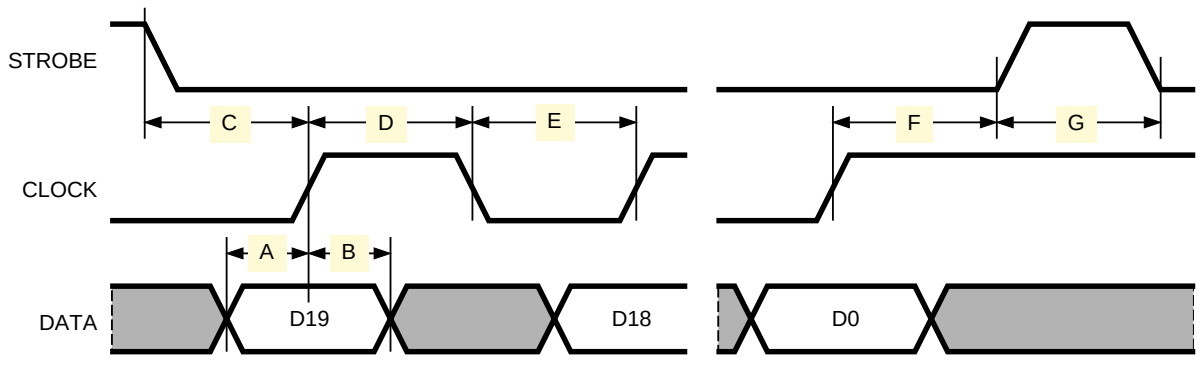
D18 Test Mode. Bit D18 low (default) operates the device in normal mode. D18 is only used for testing purposes. The user should never change this bit.

D19 Sleep Mode. Bit D19 selects a Sleep mode to minimize power consumption when not in use. This disables much of the internal circuitry including the regulator and charge pump. On power up the serial port is initialized to all 0s. Bit D19 should be programmed high for 1 ms before attempting to enable any output driver.

D19	Sleep Mode
0	Sleep
1	Normal

Serial Port Write Timing Operation. Data is clocked into the shift register on the rising edge of the CLOCK signal. Normally STROBE will be held high, only brought low to initiate a write cycle. Refer to diagram below and the specification table below for the minimum timing requirements.

A.DATA setup time ..... 15 ns
B.DATA hold time ..... 10 ns
C.Setup STROBE to CLOCK rising edge ..... 50 ns
D.CLOCK high pulse width ..... 50 ns
E.CLOCK low pulse width ..... 50 ns
F.Setup CLOCK rising edge to STROBE ..... 50 ns
G.STROBE pulse width ..... 50 ns



Dwg. WP-038

Serial Port Write Timing

FUNCTIONAL DESCRIPTION (continued)

V_{REG}. This internally generated voltage is used to operate the sink-side DMOS outputs. The V_{REG} terminal should be decoupled with a 0.22 μF capacitor to ground. V_{REG} is internally monitored and in the case of a fault condition, the outputs of the device are disabled.

Charge Pump. The charge pump is used to generate a gate-supply voltage greater than V_{BB} to drive the source-side DMOS gates. A 0.22 μF ceramic capacitor should be connected between CP_1 and CP_2 for pumping purposes. A 0.22 μF ceramic capacitor should be connected between CP and V_{BB} to act as a reservoir to operate the high-side DMOS devices. The CP voltage is internally monitored and, in the case of a fault condition, the source outputs of the device are disabled.

Shutdown. In the event of a fault (excessive junction temperature, or low voltage on CP or V_{REG}) the outputs of the device are disabled until the fault condition is removed. At power up, and in the event of low V_{DD}, the UVLO circuit disables the drivers and resets the data in the serial port to all zeros.

PWM Timer Function. The PWM timer is programmable via the serial port (bits D2 – D10) to provide off-time PWM signals to the control circuitry. In the mixed current-decay mode, the first portion of the off time operates in fast decay, until the fast decay time count (serial bits D7 – D10) is reached, followed by slow decay for the rest of the off-time period (bits D2 – D6). If the fast decay time is set longer than the off time the device effectively operates in fast decay mode. Bit D17, in conjunction with MODE, selects mixed or slow decay.

PWM Blank Timer. When a source driver turns ON, a current spike occurs due to the reverse recovery currents of the clamp diodes and/or switching transients related to distributed capacitance in the load. To prevent this current spike from erroneously resetting the source-enable latch, the sense comparator is blanked. The blank timer runs after the off-time counter (see bits D2 – D6) to provide the programmable blanking function. The blank timer is reset when ENABLE is chopped or PHASE is changed. For external PWM control, a PHASE change or ENABLE ON will trigger the blanking function.

Synchronous Rectification. When a PWM off cycle is triggered, either by an ENABLE chop command or internal fixed off-time cycle, load current will recirculate according to the decay mode selected by the control logic. The A3958 synchronous rectification feature will turn ON the opposite pair of DMOS outputs during the current decay and effectively short out the body diodes with the low $r_{DS(on)}$ driver. This will reduce power dissipation significantly and can eliminate the need for external Schottky diodes.

Synchronous rectification can be configured in active mode, passive mode, or disabled via the serial port (bits D11 and D12).

The active or passive mode selection has no impact in slow-decay mode. With synchronous rectification enabled, the slow-decay mode serves as an effective brake mode.

Current Regulation. Load current is regulated by an internal fixed off-time PWM control circuit. When the outputs of the DMOS H bridge are turned ON, the current increases in the motor winding until it reaches a trip value determined by the external sense resistor (R_S), the applied analog reference voltage (V_{REF}), the RANGE logic level, and serial data bit D16:

When RANGE = D16 ..... $I_{TRIP} = V_{REF}/10R_S$
When RANGE ≠ D16 ..... $I_{TRIP} = V_{REF}/5R_S$

At the trip point, the sense comparator resets the source-enable latch, turning OFF the source driver. The load inductance then causes the current to recirculate for the serial-port-programmed fixed off-time period. The current path during recirculation is determined by the configuration of slow/mixed current-decay mode (D17) and the synchronous rectification control bits (D11 and D12).

APPLICATIONS INFORMATION

Current Sensing. To minimize inaccuracies in sensing the I_{TRIP} current level, which may be caused by ground trace I•R drops, the sense resistor should have an independent ground return to the ground terminal of the device. For low-value sense resistors the I•R drops in the PCB sense resistor's traces can be significant and should be taken into account. The use of sockets should be avoided as they can introduce variation in R_S due to their contact resistance.

The maximum value of R_S is given as $R_S \leq 0.5/I_{TRIP}$.

Braking. The braking function is implemented by driving the device in slow-decay mode via serial port bit D13, enabling synchronous rectification via bit D12, andchopping with the combination of D14 and the ENABLE input terminal. Because it is possible to drive current in either direction through the DMOS drivers, this configuration effectively shorts out the motor-generated BEMF as long as the ENABLE chop mode is asserted. It is important to note that the internal PWM current-control circuit will not limit the current when braking, because the current does not flow through the sense resistor. The maximum brake current can be approximated by V_{BEMF}/R_L. Care should be taken to ensure that the maximum ratings of the device are not exceeded in worst-case braking situations of high speed and high inertial loads.

Thermal Protection. Circuitry turns OFF all drivers when the junction temperature reaches 165°C typically. It is intended only to protect the device from failures due to excessive junction temperatures and should not imply that output short circuits are permitted. Thermal shutdown has a hysteresis of approximately 15°C.

Layout. The printed wiring board should use a heavy ground plane. For optimum electrical and thermal performance, the driver should be soldered directly onto the board. The ground side of R_S should have an individual path to the ground terminals of the device. This path should be as short as is possible physically and should not have any other components connected to it. It is recommended that a 0.1 µF capacitor be placed between SENSE and ground as close to the device as possible; the load supply terminal, V_{BB}, should be decoupled with an electrolytic capacitor (> 47 µF is recommended) placed as close to the device as is possible.

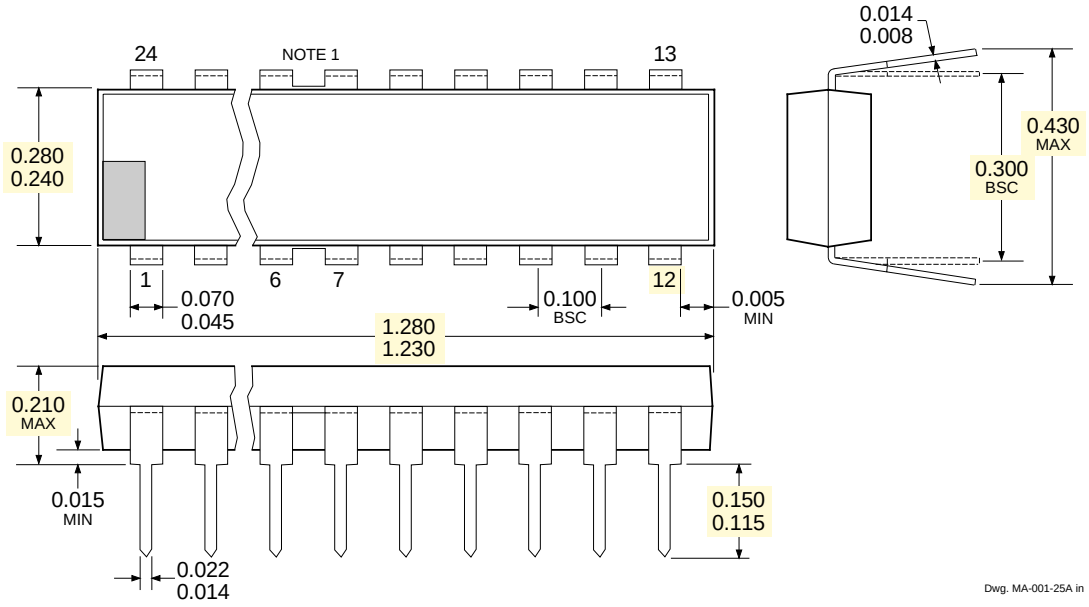
Terminal List

Terminal Name	Terminal Description	A3958SLB (SOIC)	A3958SB (DIP)
CP	Reservoir capacitor (typically 0.22 μF)	1	24
$CP_{1\ \&}\ CP_2$	The charge pump capacitor (typically 0.22 μF)	2 & 3	1 & 2
PHASE	Logic input for direction control (see also D15)	4	3
OSC	Logic-level oscillator (square wave) input	5	4
GROUND	Grounds	6, 7	5, 6, 7, 8*
LOGIC SUPPLY	V_{DD}, the low voltage (typically 5 V) supply	8	9
ENABLE	Logic input for enable control (see also D14)	9	10
DATA	Logic-level input for serial interface	10	11
CLOCK	Logic input for serial port (data is entered on rising edge)	11	12
STROBE	Logic input for serial port (active on rising edge)	12	13
REF	V_{REF}, the load current reference input volt. (see also D16)	13	14
MODE	Logic input for PWM mode control (see also D17)	14	15
NO CONNECT	No (Internal) Connection	15	—
OUT_A	One of two DMOS bridge outputs to the motor	16	16
SENSE	Sense resistor	17	17
GROUND	Grounds	18, 19	18, 19
LOAD SUPPLY	V_{BB}, the high-current, 20 V to 50 V, motor supply	20	20
OUT_B	One of two DMOS bridge outputs to the motor	21	21
NO CONNECT	No (Internal) connection	22	—
RANGE	Logic Input for V_{REF} range control (see also D16)	23	22
V_{REG}	Regulator decoupling capacitor (typically 0.22 μF)	24	23

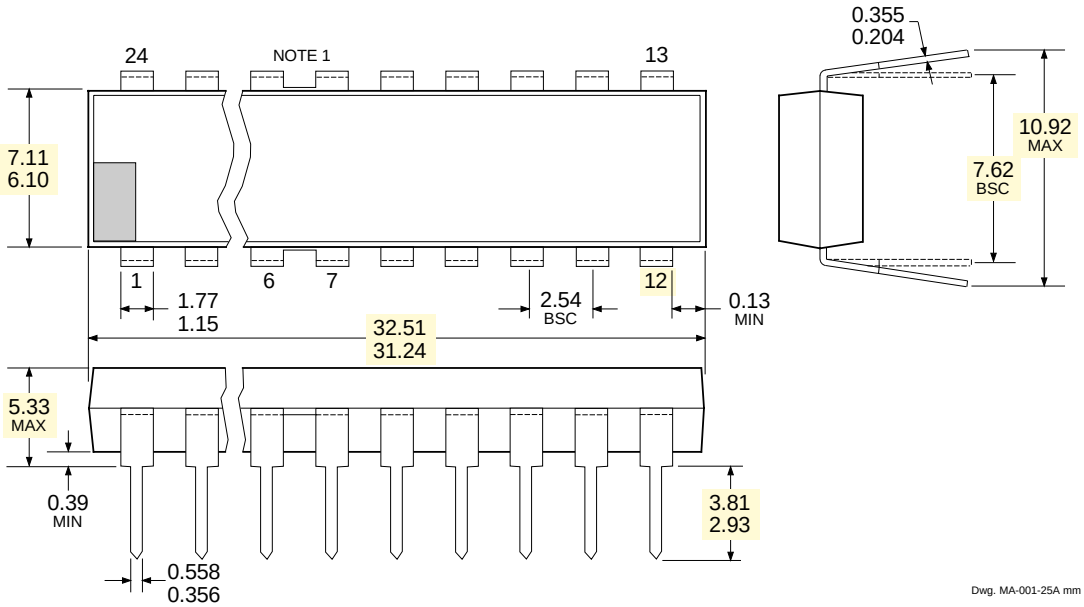
* For the A3958SB DIP only, there is an indeterminate resistance between the substrate grounds (pins 6, 7, 18, and 19) and the grounds at pins 5 and 8. Pins 5 and 8, and 6, 7, 18, or 19 must be connected together externally.

A3958SB

Dimensions in Inches
(controlling dimensions)



Dimensions in Millimeters
(for reference only)

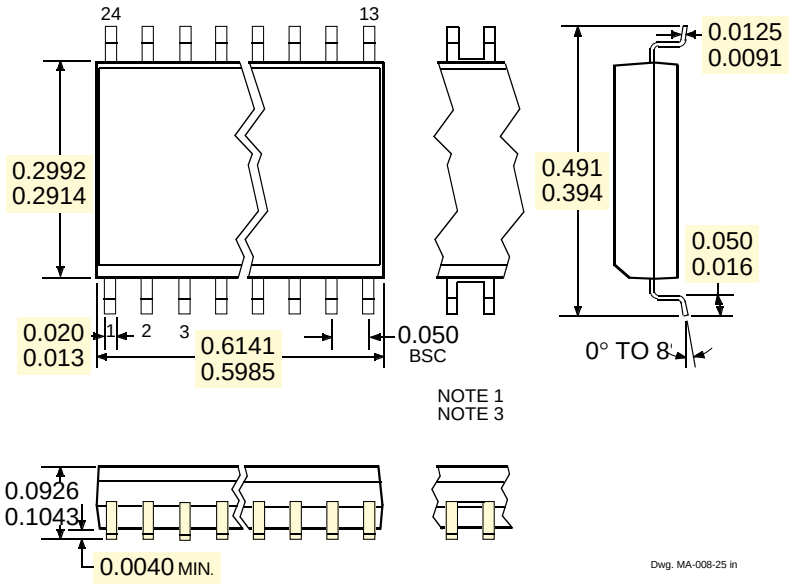


NOTES:
1. Webbed lead frame. Leads 6, 7, 18, and 19 are internally one piece.
2. Exact body and lead configuration at vendor's option within limits shown.
3. Lead spacing tolerance is non-cumulative.
4. Lead thickness is measured at seating plane or below.

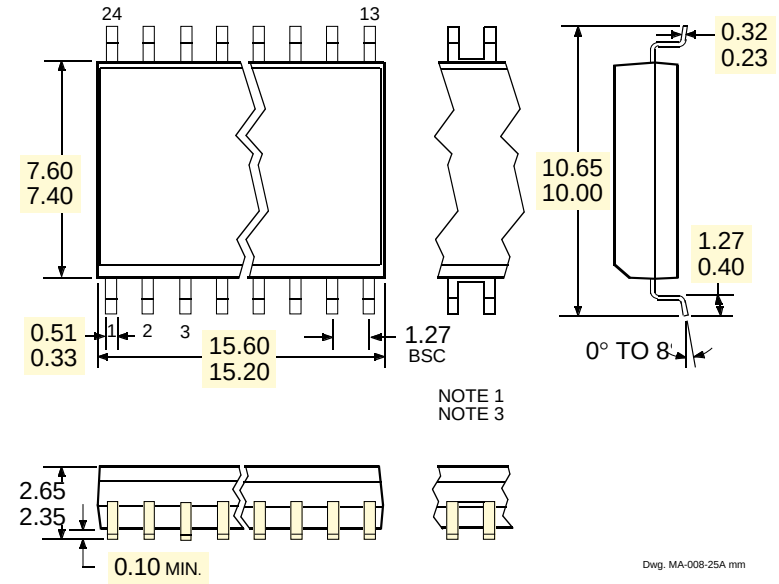
Allegro® MicroSystems, Inc.
115 Northeast Cutoff, Box 15036
Worcester, Massachusetts 01615-0036 (508) 853-5000

A3958SLB

Dimensions in Inches
(for reference only)



Dimensions in Millimeters
(controlling dimensions)



NOTES: 1. Exact body and lead configuration at vendor's option within limits shown.
2. Lead spacing tolerance is non-cumulative.
3. Webbed lead frame. Leads 6, 7, 18, and 19 are internally one piece.

3958
DMOS FULL-BRIDGE
PWM MOTOR DRIVER

MOTOR DRIVERS SELECTION GUIDE

Function	Output Ratings *		Part Number †
INTEGRATED CIRCUITS FOR BRUSHLESS DC MOTORS			
3-Phase Controller/Drivers	±2.0 A	45 V	2936 & 2936-120
Hall-Effect Latched Sensors	10 mA	24 V	3175 & 3177
2-Phase Hall-Effect Sensor/Controller	20 mA	25 V	3235
Hall-Effect Complementary-Output Sensor	20 mA	25 V	3275
2-Phase Hall-Effect Sensor/Driver	900 mA	14 V	3625
2-Phase Hall-Effect Sensor/Driver	400 mA	26 V	3626
3-Phase Power MOSFET Controller	—	28 V	3933
Hall-Effect Complementary-Output Sensor/Driver	300 mA	60 V	5275
3-Phase Back-EMF Controller/Driver	±900 mA	14 V	8902–A
3-Phase Back-EMF Controller/Driver	±1.0 A	7 V	8984
INTEGRATED BRIDGE DRIVERS FOR DC AND BIPOLAR STEPPER MOTORS			
PWM Current-Controlled Dual Full Bridge	±750 mA	45 V	2916
PWM Current-Controlled Dual Full Bridge	±1.5 A	45 V	2917
PWM Current-Controlled Dual Full Bridge	±1.5 A	45 V	2918
PWM Current-Controlled Dual Full Bridge	±750 mA	45 V	2919
Dual Full-Bridge Driver	±2.0 A	50 V	2998
PWM Current-Controlled Full Bridge	±2.0 A	50 V	3952
PWM Current-Controlled Full Bridge	±1.3 A	50 V	3953
PWM Current-Controlled Microstepping Full Bridge	±1.5 A	50 V	3955
PWM Current-Controlled Microstepping Full Bridge	±1.5 A	50 V	3957
DMOS Full Bridge PWM Driver	±2.0 A	50 V	3958
PWM Current-Controlled Dual Full Bridge	±800 mA	33 V	3964
PWM Current-Controlled Dual Full Bridge	±650 mA	30 V	3966
PWM Current-Controlled Dual Full Bridge	±650 mA	30 V	3968
PWM Current-Controlled Dual Full Bridge	±750 mA	45 V	6219
OTHER INTEGRATED CIRCUIT & PMCM MOTOR DRIVERS			
Unipolar Stepper-Motor Quad Driver	1.8 A	50 V	2544
Unipolar Stepper-Motor Translator/Driver	1.25 A	50 V	5804
Unipolar Stepper-Motor Quad Drivers	1 A	46 V	7024 & 7029
Unipolar Stepper-Motor Quad Drivers	3 A	46 V	7026
Unipolar Microstepper-Motor Quad Driver	1.2 A	46 V	7042
Unipolar Microstepper-Motor Quad Driver	3 A	46 V	7044
Voice-Coil Motor Driver	±500 mA	6 V	8932–A
Voice-Coil Motor Driver	±800 mA	16 V	8958
Voice-Coil (and Spindle) Motor Driver	±350 mA	7 V	8984

* Current is maximum specified test condition, voltage is maximum rating. See specification for sustaining voltage limits or over-current protection voltage limits. Negative current is defined as coming out of (sourcing) the output.

† Complete part number includes additional characters to indicate operating temperature range and package style.

Allegro® MicroSystems, Inc.
115 Northeast Cutoff, Box 15036
Worcester, Massachusetts 01615-0036 (508) 853-5000