

# **STK4036XI**

# AF Power Amplifier (Split Power Supply) (50W min, THD = 0.008%)

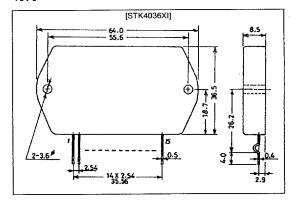
#### **Features**

- · Compact packaging supports slimmer set designs
- Series designed from 50 up to 150 W and pincompatibility
- Simpler heat sink design facilitates thermal design of slim stereo sets
- Current mirror circuit, cascade circuit and purecomplimentary circuit application reduce distortion to 0.008 %
- Supports addition of electronic circuits for thermal shutdown and load-short protection circuit as well as pop noise muting which occurs when the power supply switch is turned on and off.

## **Package Dimensions**

unit: mm

4075



#### **Specifications**

#### Maximum Ratings at Ta = 25°C

Parameter	Symbol	Condition	Rating	Unit
Maximum supply voltage	V <sub>CC</sub> max		± 53.5	V
Thermal resistance	θj-c		1.8	°C/W
Junction temperature	Tj		150	°C
Operating substrate temperature	Tc		125	°C
Storage temperature	Tstg		-30 to +125	°C
Available time for load shorted	t <sub>s</sub> *1	$V_{CC} = \pm 37 \text{ V}, R_1 = 8 \Omega, f = 50 \text{ Hz}, P_O = 50 \text{ W}$	. 1	s

#### Recommended Operational Conditions at Ta = 25°C

Parameter	Symbol	Condition	Rating	Unit
Recommended supply voltage	Vcc		± 37	V
Load resistance	RL		8	Ω

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#### **Operating Characteristics**

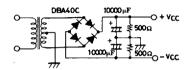
at Ta = 25°C,  $V_{CC}$  =  $\pm$  37 V,  $R_L$  = 8  $\Omega$ , VG = 40 dB, Rg = 600  $\Omega$ , 100 k LPF ON,  $R_L$  (non-inductive)

Parameter		Condition	Rating			
	Symbol		min	typ	max	Unit
Quiescent current	lcco	V <sub>CC</sub> = ± 44.5 V	15		120	mA
Output power	Po	THD = 0.008 %, f = 20 Hz to 20 kHz	50			W
Total harmonic distortion	THD	P <sub>O</sub> = 1.0 W, f = 1 kHz			0.008	%
Frequency response	fL, fH	$P_0 = 1.0 \text{ W}, + 0 \text{ dB}$		20 to 50k		Hz
Input resistance	ri	P <sub>O</sub> = 1.0 W, f = 1 kHz		55		kΩ
Output noise voltage	V <sub>NO</sub> *2	V <sub>CC</sub> = ± 44.5 V, Rg = 10 kΩ			1.2	mVrms
Neutral voltage	V <sub>N</sub>	V <sub>CC</sub> = ± 44.5 V	-70	0	+ 70	m∨

Note: Use rated power supply for test unless otherwise specified.

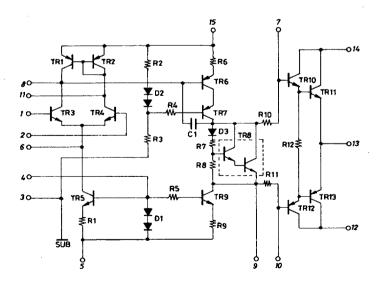
\*1 When measuring permissible load short time and output noise voltage use transformer power supply indicated below.

<sup>\*2</sup> Output noise voltage represents the peak value on the rms scale (VTVM). The noise voltage waveform does not include the pulse noise.

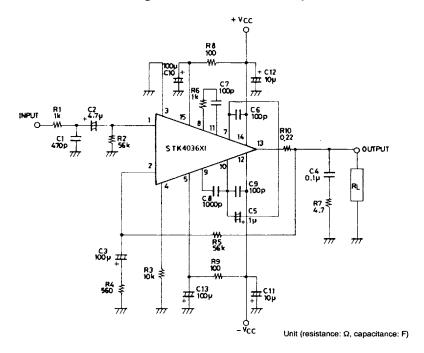


**Specified Transformer Power Supply** (MG-200 Equivalent)

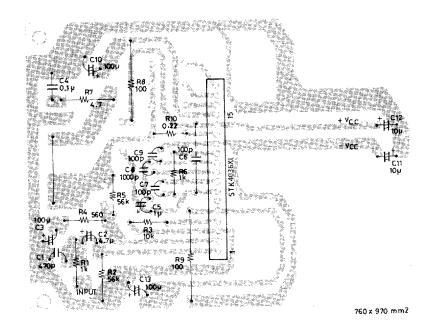
#### **Equivalent Circuit**



### Application Circuit: 50W min Single Channel AF Power Amplifier



# Sample Printed Circuit Pattern for Application Circuit (Copper-foiled side)



Unit (resistance:  $\Omega$ , capacitance: F)

## **Description of External Parts**

R<sub>1</sub>, C<sub>1</sub> : Input filter circuit

• Reduces high-frequency noise.

C<sub>2</sub>: Input coupling capacitor

 DC current suppression. A reduction in reactance is effective because of increases in capacitor reactance at low frequencies and 1/f noise dependence on signal source resistance which result in output noise worsening.

C<sub>3</sub> : AC NF capacitor R<sub>4</sub>, R<sub>5</sub> : Used for VG setting.

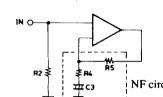
R<sub>2</sub>: Input bias resistor

· Biases the input pin to zero.

• Effects V<sub>N</sub> stability (refer to NF circuit).

• Due to differential input, input resistance is more or less determined by this resistance value.

R<sub>4</sub>, R<sub>5</sub> : NFB circuit (AC NF circuit). Use of resistor with 1% error is suggested. C<sub>3</sub> (R<sub>2</sub>)



• VG settings are obtained using R<sub>4</sub> and R<sub>5</sub> according to the following equation:

 $\log_{20} \frac{R_5}{R_4}$  40 dB is recommended.

• Low-frequency cutoff frequency settings are obtained using R<sub>4</sub> and C<sub>3</sub> according to the following equation:

$$f_L = \frac{1}{2\pi \cdot R_4 \cdot C_3} \quad [Hz]$$

When changing the VG setting, you should change  $R_4$  which requires a recheck of the low cutoff frequency setting. When the VG setting is changed using  $R_5$ , the setting should ensure  $R_2$  equals  $R_5$  so that  $V_N$  balance stability is maintained. If the resistor value is increased more than the existing value,  $V_N$  balance may be disturbed and result in deterioration of  $V_N$  temperature characteristics.

R<sub>3</sub>: Differential constant-current bias resistor

R<sub>6</sub>, R<sub>7</sub> : For oscillation suppression and phase compensation applications

(For use with differential stage applications)

R<sub>7</sub>, C<sub>4</sub>: For oscillation suppression and phase compensation applications

(A Mylar capacitor is recommended for C<sub>4</sub> for use with output stage applications)

C<sub>6</sub>, C<sub>9</sub>: For oscillation suppression and phase compensation applications

Power stage (Must be connected near the pin) C<sub>6</sub>: Positive (+) power C<sub>9</sub>: Negative (-) power

C<sub>8</sub>: For oscillation suppression and phase compensation applications

(Oscillation suppression before power step clip)

C<sub>5</sub>: For oscillation suppression and distortion improvement applications

 $R_8, C_{10}$  : Ripple filter circuit on positive (+) side.  $R_9, C_{13}$  : Ripple filter circuit on negative (-) side.  $C_{11}, C_{12}$  : For oscillation suppression applications

• Used for reducing power supply impedance to stable IC operation and should be connected near the IC

pin. We recommend that you use an electrolytic capacitor.

R<sub>10</sub> : Output resistor
Increases load shorting endurance capacity during times of high output.